

# **New Developments in Tail-Equivalent Linearization method for Nonlinear Stochastic Dynamics**

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**Acknowledgement:  
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# The American University of Armenia



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# Motivation

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- ❑ Stochasticity and nonlinearity are essential considerations in assessing the reliability of structural and mechanical systems under extreme loads, e.g.
  - Inelastic response to earthquake ground motion
  - Response to wave loading under material and/or geometric nonlinearities
  - Response to turbulent wind
- ❑ Existing methods of nonlinear stochastic dynamic analysis are restricted to special cases, or are not well suited for reliability analysis – hence the need for a new method.

# Outline

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- ❑ Methods for nonlinear stochastic dynamic analysis
- ❑ The equivalent linearization method (ELM)
- ❑ Basic elements of TELM:
  - Characterization of linear systems
  - The first-order reliability method (FORM)
  - Discretization of stochastic excitation
- ❑ The tail-equivalent linearization method (TELM)
- ❑ Characteristics of TELM
- ❑ Applications of TELM
- ❑ Challenges and limitation of TELM
- ❑ Concluding remarks

# Methods for nonlinear stochastic dynamic analysis

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- ❑ Classical methods
  - Fokker-Planck equation
  - Stochastic averaging
  - Moment/cumulant closure
  - Perturbation
- ❑ The equivalent linearization method (ELM)
- ❑ Monte Carlo simulation

# Equivalent linearization methods (ELM)

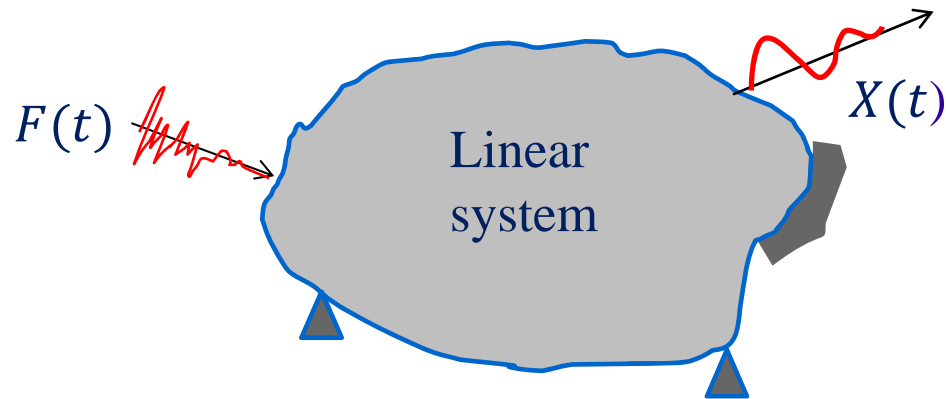
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- ❑ Approximates the nonlinear response in terms of the response of an “equivalent” linear system (Caughey 1963).
- ❑ The ELS is determined by minimizing a measure of discrepancy between nonlinear and linear responses:
  - Conventional ELM – minimize the variance of the error between nonlinear and linear responses; requires the assumption of a distribution, typically Gaussian (e.g., Atalik & Utku 1976; Wen 1976).
  - Minimize the difference in mean up-crossing rates at a selected threshold (Casciati et al. 1993).
  - Minimize higher moments of the error (Naess 1995).
  - TELM – set the tail probability of the linear response equal to the first-order approximation of the tail probability of the nonlinear response (Fujimura and Der Kiureghian 2007).



# Characterization of a linear system

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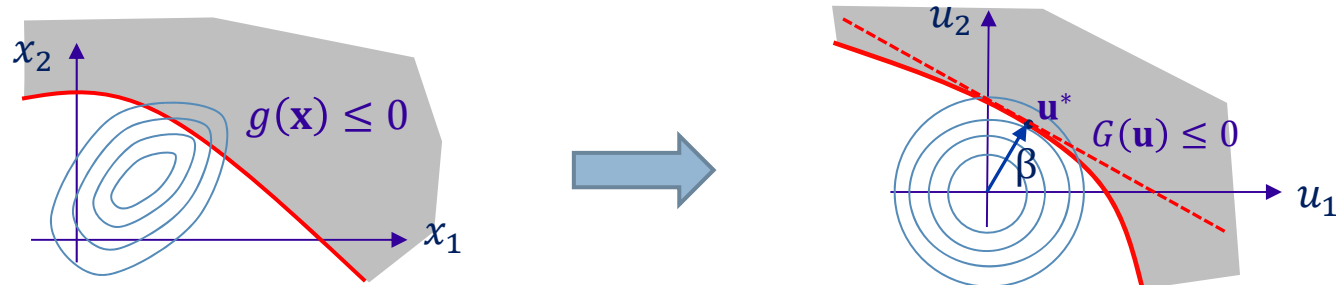
- For an input-output pair  $(F(t), X(t))$ , a stable linear system is completely defined by either of the following:
  - $h(t)$  = impulse response function (IRF), i.e., response to  $F(t) = \delta(t)$
  - $H(\omega)$  = frequency response function (FRF), i.e., amplitude of steady-state response to  $F(t) = \exp(i\omega t)$

# The first-order reliability method (FORM)

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- An approximate method for solving time-invariant reliability problems:

- $\mathbf{x}$  = vector of random variables  
 $g(\mathbf{x})$  = limit-state function:  $\{g(\mathbf{x}) \leq 0\}$  = failure event  
 $p_f = \Pr[g(\mathbf{x}) \leq 0]$  probability of failure



- $\mathbf{u} = \mathbf{u}(\mathbf{x})$  transformation to standard normal space  
 $G(\mathbf{u}) = g(\mathbf{x}(\mathbf{u}))$  limit-state function in transformed space  
 $\mathbf{u}^* = \min \arg\{\|\mathbf{u}\| \mid G(\mathbf{u}) = 0\}$  design point  
 $\beta = \|\mathbf{u}^*\|$  reliability index  
 $p_f \approx \Phi(-\beta)$ , FORM approximation



# Discrete representation of a stochastic process

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- General form for a zero-mean Gaussian process

$$F(t) = \mathbf{s}(t) \cdot \mathbf{u}$$

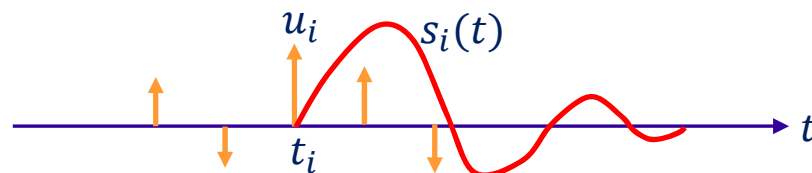
$\mathbf{s}(t) = [s_1(t) \cdots s_n(t)]$  vector of deterministic basis functions

$\mathbf{u} = [u_1 \cdots u_n]$  vector of standard normal random variables

- Time-domain discretization (modulated filtered white noise)

$$s_i(t) = q(t)h_f(t - t_i)$$

$h_f(\cdot)$  = impulse response function of a linear filter



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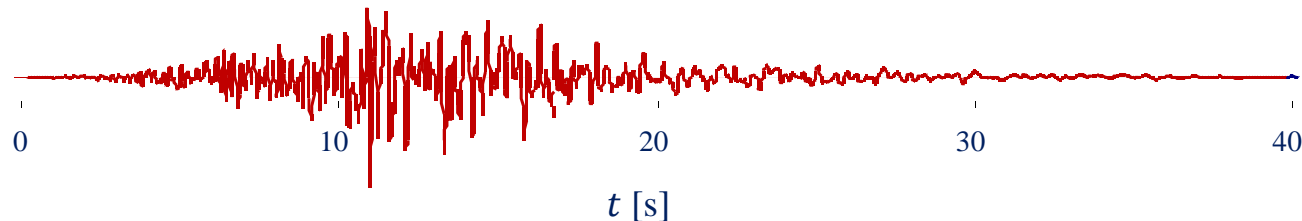
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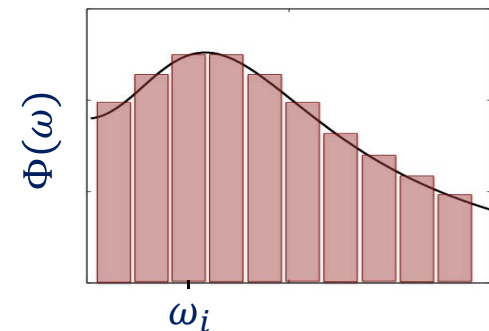
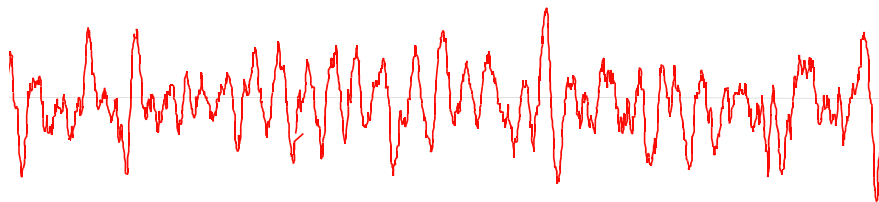
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- Frequency-domain discretization (stationary process)

$$F(t) = \sum_{i=1}^{n/2} \sigma_i [u_i \sin(\omega_i t) + \bar{u}_i \cos(\omega_i t)]$$

$$s_i(t) = \sigma_i \sin(\omega_i t), \quad \bar{s}_i(t) = \sigma_i \cos(\omega_i t)$$



# FORM solution of stochastic dynamic problems

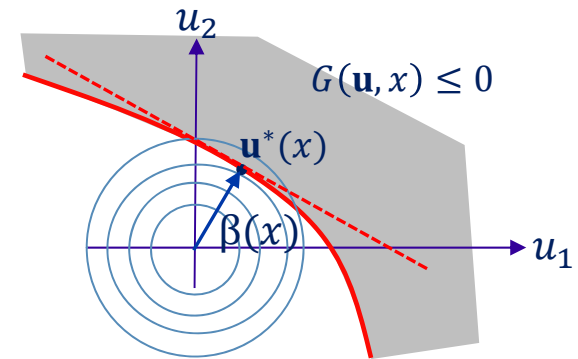
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## □ Definitions:

- $F(t) = \mathbf{s}(t) \cdot \mathbf{u}$  discretized stochastic excitation
- $X(t, \mathbf{u}) =$  response to discretized stochastic excitation
- $\Pr(x < X(t, \mathbf{u})) =$  tail probability for threshold  $x$  at time  $t$

## □ Reliability formulation:

- $G(\mathbf{u}, x) = x - X(t, \mathbf{u})$
- $\Pr(x < X(t, \mathbf{u})) = \Pr(G(\mathbf{u}, x) \leq 0)$
- $\mathbf{u}^*(x) = \arg \min\{\|\mathbf{u}\| \mid G(\mathbf{u}, x) = 0\}$
- $\beta(x) = \|\mathbf{u}^*(x)\|$
- $\Pr(x < X(t, \mathbf{u})) \approx \Phi(-\beta(x))$  FORM approx. of tail probability



# FORM solution of stochastic dynamic problems

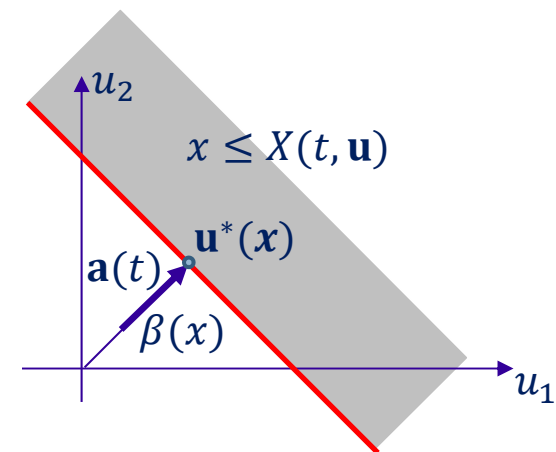
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## □ Reliability formulation – the case of linear system

- $X(t, \mathbf{u}) = \mathbf{a}(t) \cdot \mathbf{u}$ ,  $a_i(t) =$  response to  $s_i(t)$
- $G(\mathbf{u}, x) = x - \mathbf{a}(t) \cdot \mathbf{u}$
- $\mathbf{u}^*(x) = \frac{x\mathbf{a}(t)}{\|\mathbf{a}(t)\|^2}$
- $\beta(x) = \frac{x}{\|\mathbf{a}(t)\|}$
- $\Pr(x < X(t, \mathbf{u})) = \Phi(-\beta(x))$



# Identification of the linear system

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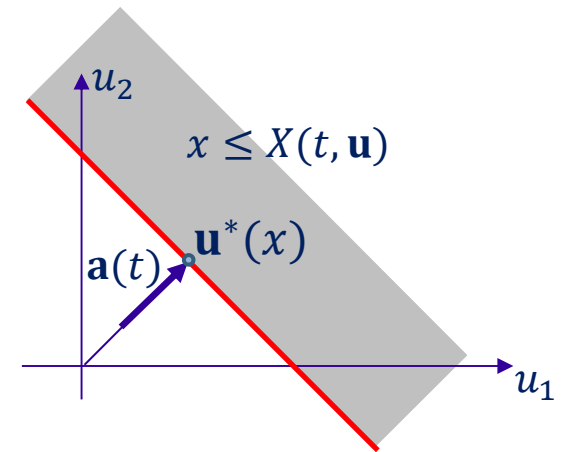
- Given the design point, one can identify the linear system (for the particular input-output pair)

$$\mathbf{u}^* \rightarrow \mathbf{a}(t) = \frac{x\mathbf{u}^*}{\|\mathbf{u}^*\|^2}$$

- Time-domain analysis:

Solve for  $h(t)$  in system of equations

$$\sum_{j=1}^n h(t - t_j) s_i(t_j) \Delta t = a_i(t), \quad i = 1, \dots, n$$



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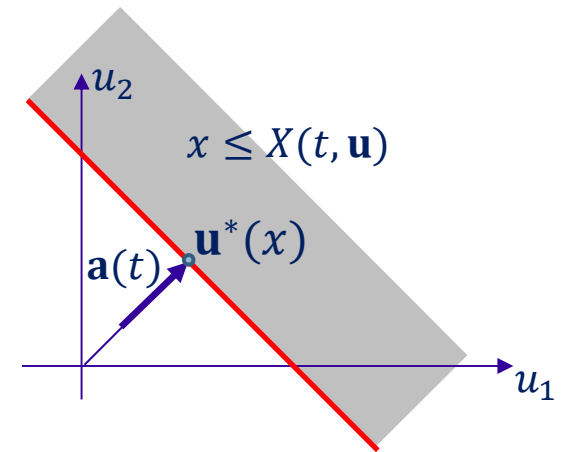
$$\mathbf{u}^* \rightarrow \mathbf{a}(t) = \frac{x\mathbf{u}^*}{\|\mathbf{u}^*\|^2}$$

- Frequency-domain analysis:

$$|H(\omega_i)| = \frac{\sqrt{a_i(t)^2 + \bar{a}_i(t)^2}}{\sigma_i}$$

$$\theta_i = \tan^{-1} \left[ \frac{a_i(t)}{\bar{a}_i(t)} \right]$$

$$H(\omega_i) = |H(\omega_i)| \exp(i\theta_i)$$

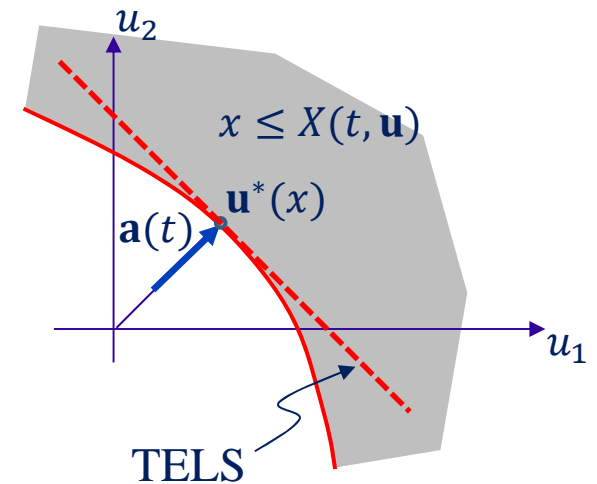




# The tail-equivalent linearization method (TELM)

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- ❑ For selected threshold  $x$  and time  $t$ , formulate the tail probability problem in terms of the limit-state function  $G(\mathbf{u}, t) = x - X(t, \mathbf{u})$
- ❑ Find the design point  $\mathbf{u}^*(x)$
- ❑ Find the gradient vector of the tangent plane  $\mathbf{a}(t) = \frac{x\mathbf{u}^*}{\|\mathbf{u}^*\|^2}$
- ❑ Identify the tail-equivalent linear system (TELS) in terms of its IRF  $h(t)$  or its FRF  $H(\omega)$



# Characteristics of TELM

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- For the given threshold  $x$  and time  $t$ ,  
the tail probability of the TELS response  
= first-order approximation of the tail probability  
of the nonlinear system response



*Tail-Equivalent Linearization Method (TELM)*

# Characteristics of TELM

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- As opposed to ELM and other linearization methods, TELM is a *non-parametric* method.



There is no need to define a parameterized linear system.

The TELS is identified numerically in terms of its IRF  $h(t)$  or its FRF  $H(\omega)$ .

# Design-point excitation and response

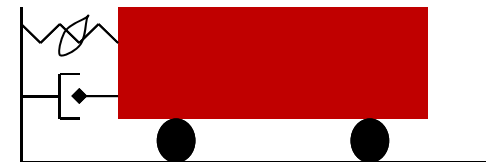
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- The design-point excitation  $F^*(t) = \mathbf{s}(t) \cdot \mathbf{u}^*$  represents the *most likely realization* of the stochastic excitation to give rise to the event  $\{x \leq X(t, \mathbf{u})\}$

- Example:

$$m\ddot{X} + c\dot{X} + k[\alpha X + (1 - \alpha)Z] = F(t)$$

$$\dot{Z} = -\gamma|\dot{X}||Z|^{n-1}Z - \eta|Z|^n\dot{X} + A\dot{X}$$

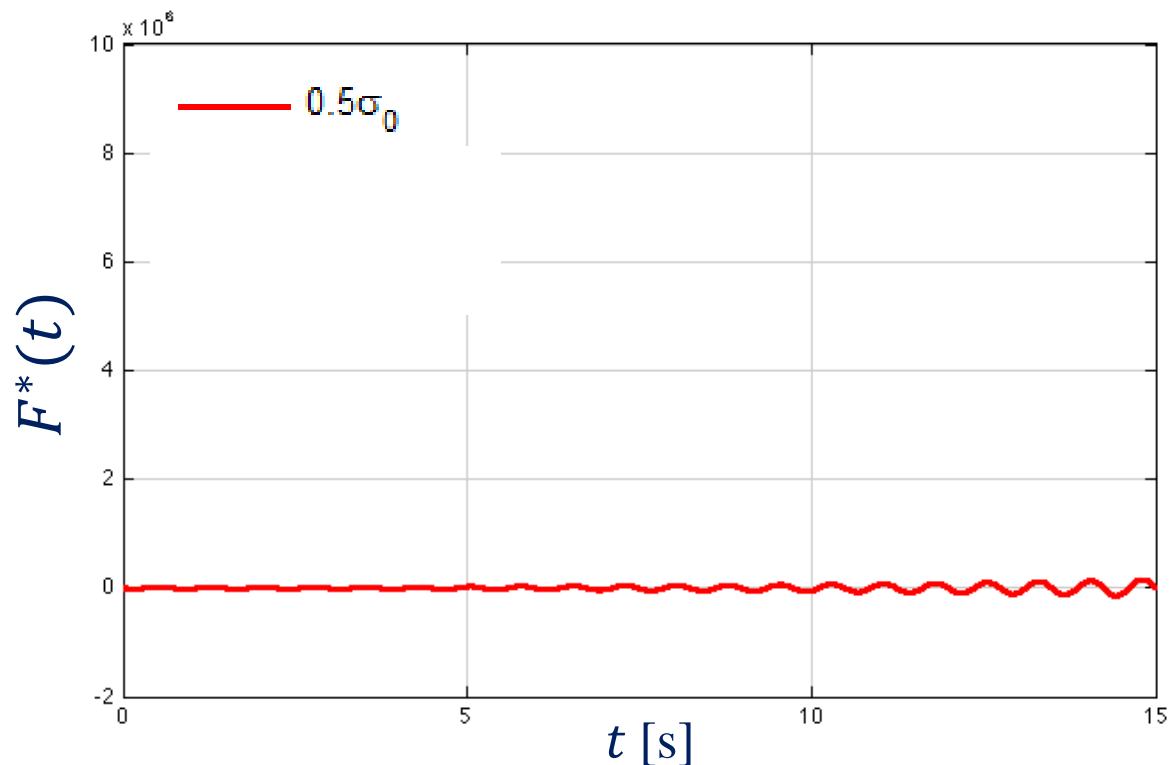


$$\begin{aligned} m &= 3E5\text{kg}; c = 251\text{kNs/m} \\ k &= 2.1E4 \text{ kN/m}, \alpha = 0.1 \\ \gamma &= \eta = 1/2\sigma_0, n = 3, A = 1 \\ F(t) &\text{ white noise, } S = 1\text{m}^2/\text{s}^3\text{rad} \\ \sigma_0 &= \pi S m^2 / ck \end{aligned}$$

# Design-point excitation and response

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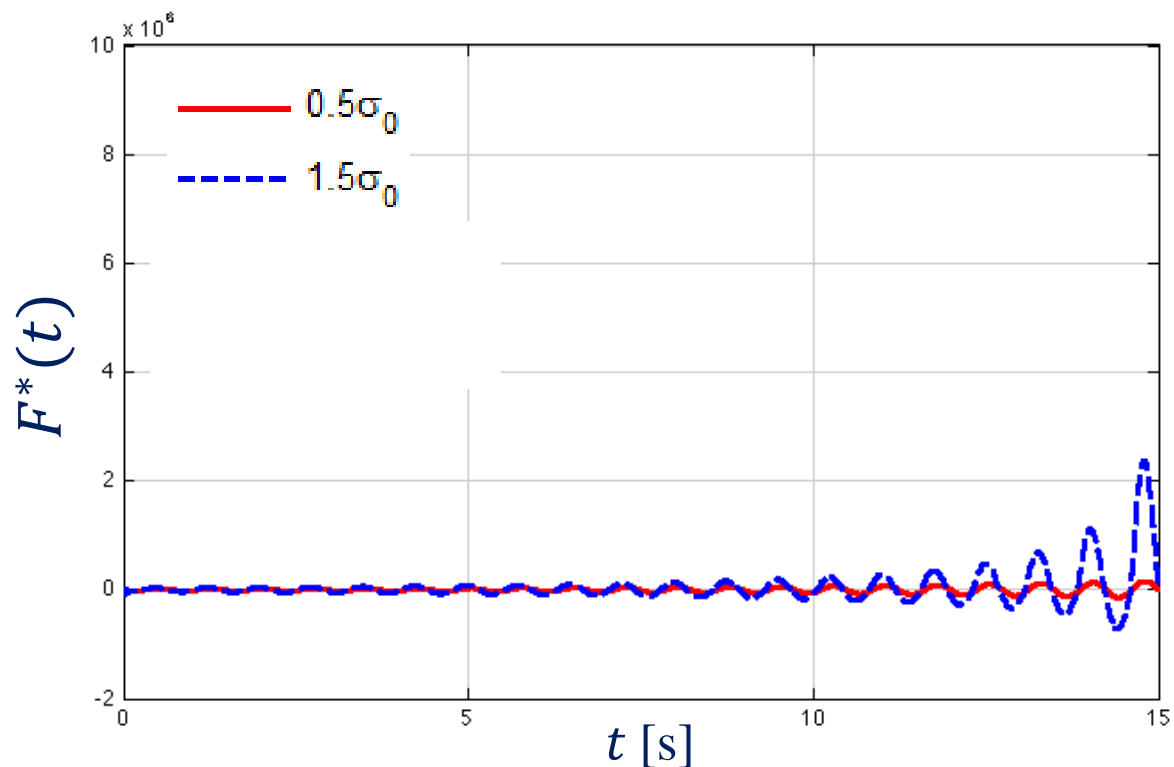
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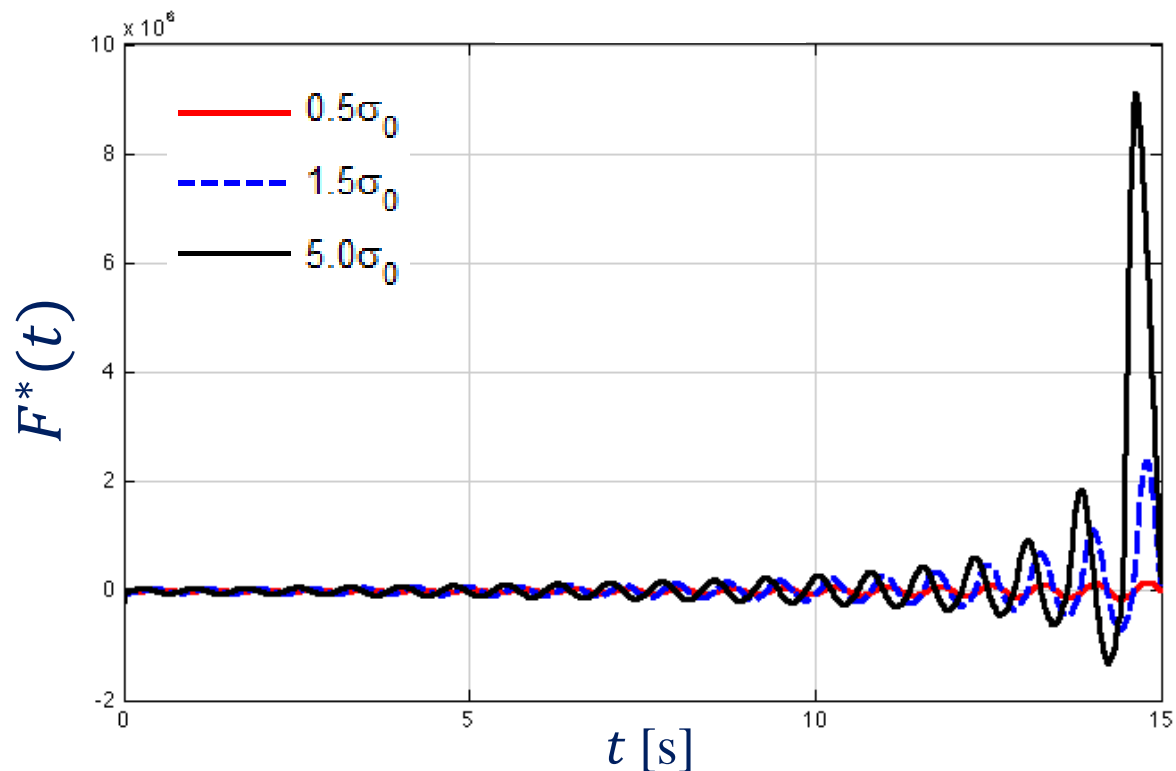
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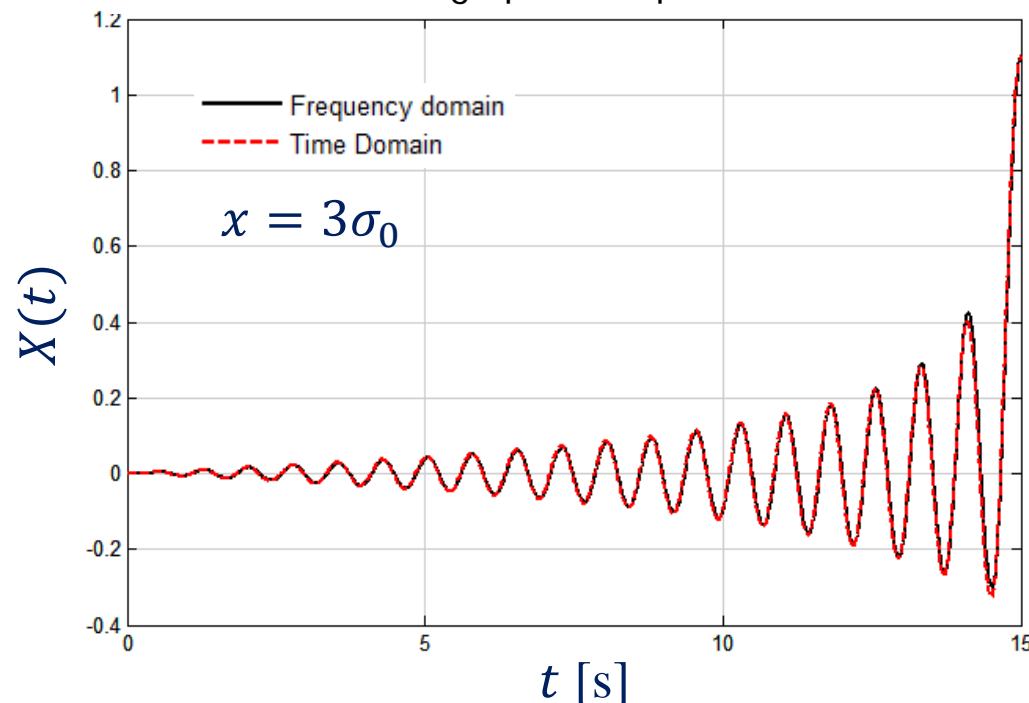


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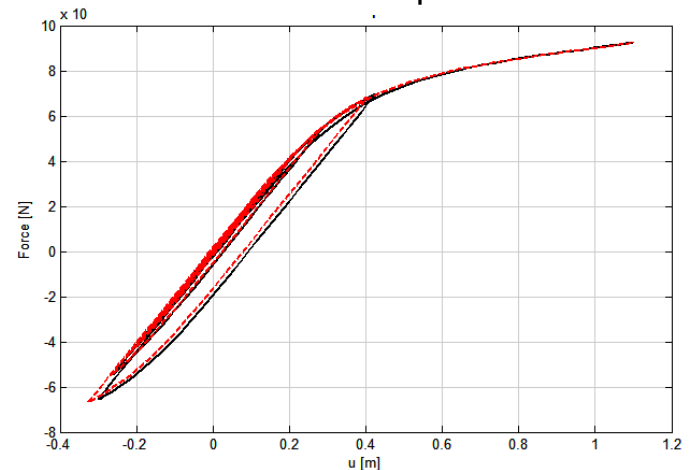
■23

- The design-point response  $X(t, \mathbf{u}^*)$  represents the *most likely realization* of the response trajectory leading to  $X(t, \mathbf{u}) = x$ .

Design point response



Force versus displacement



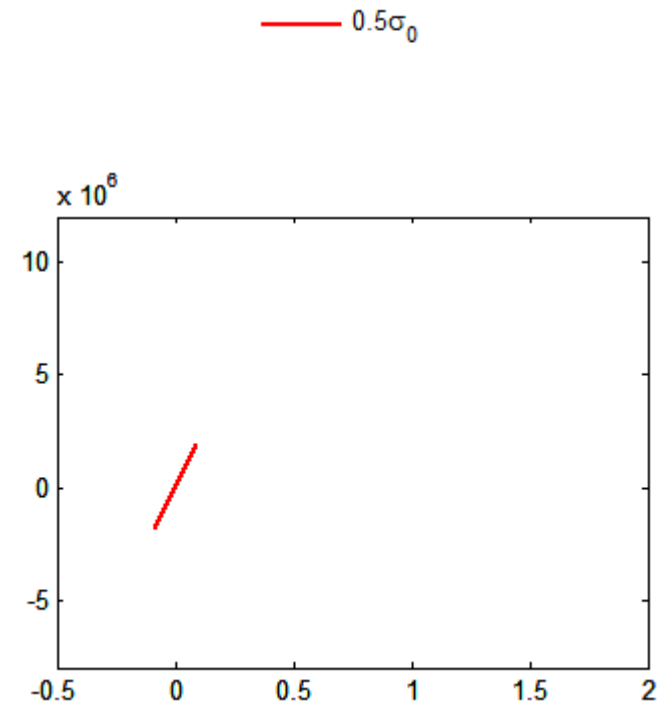
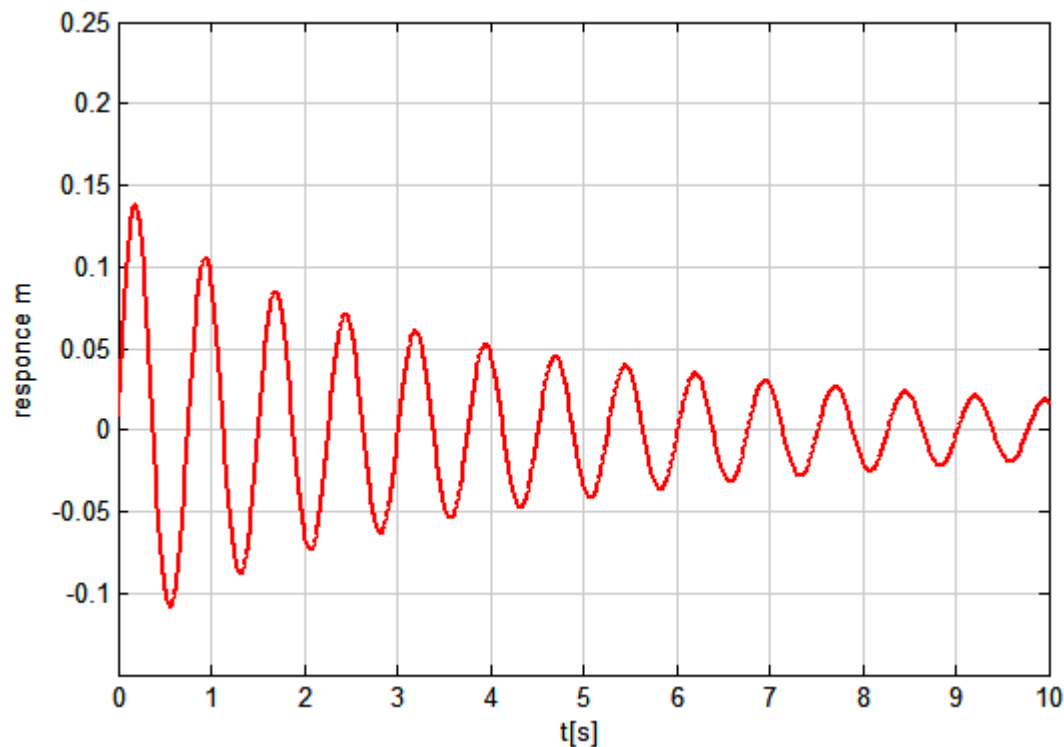
# Characteristics of TELS

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- The TELS strongly depends on the selected threshold:

$$h(t) \rightarrow h(t, x), \quad H(\omega) \rightarrow H(\omega, x)$$

IRF of the TELS for selected thresholds

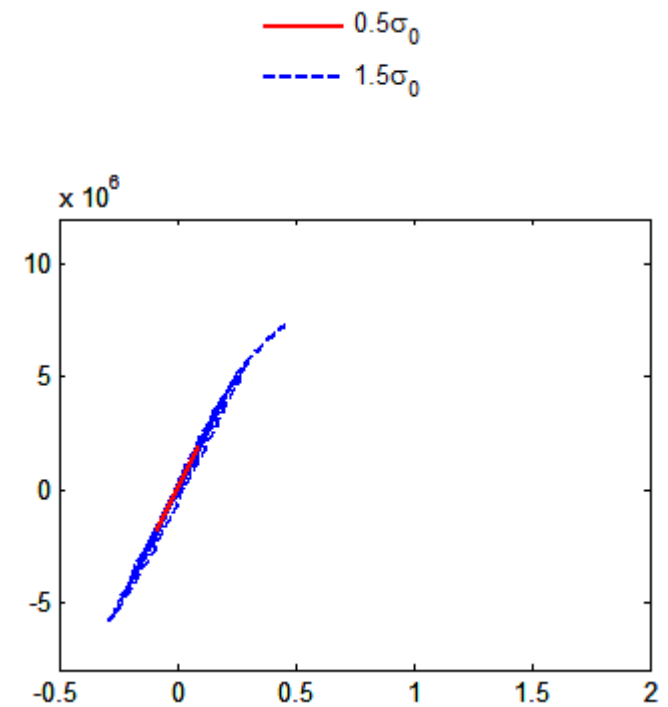
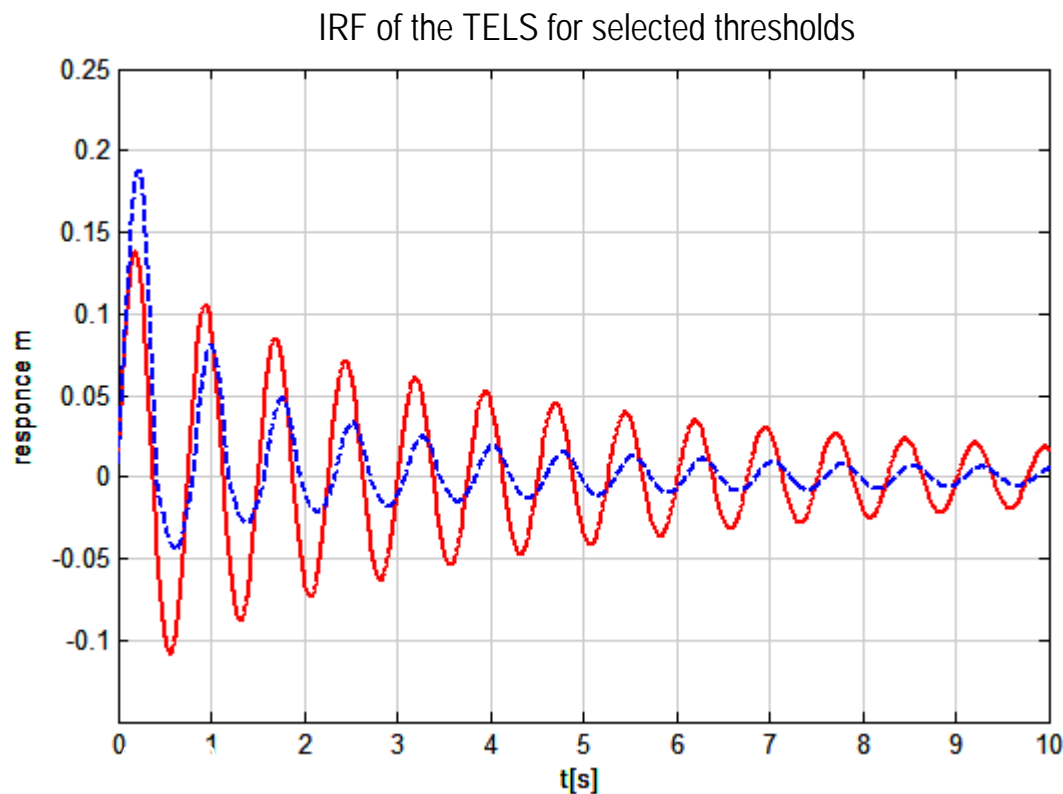


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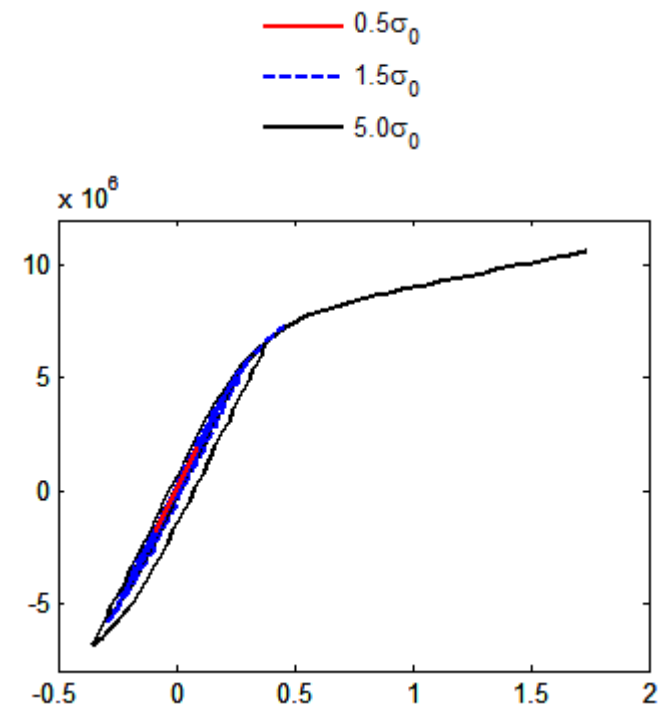
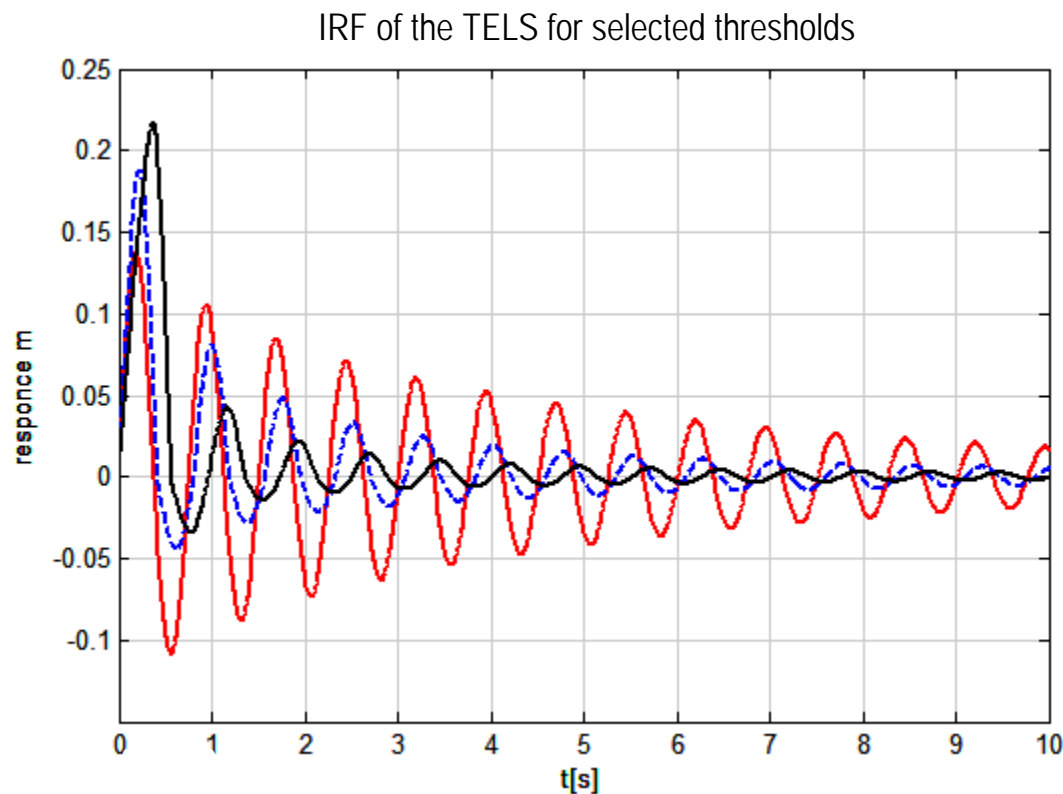


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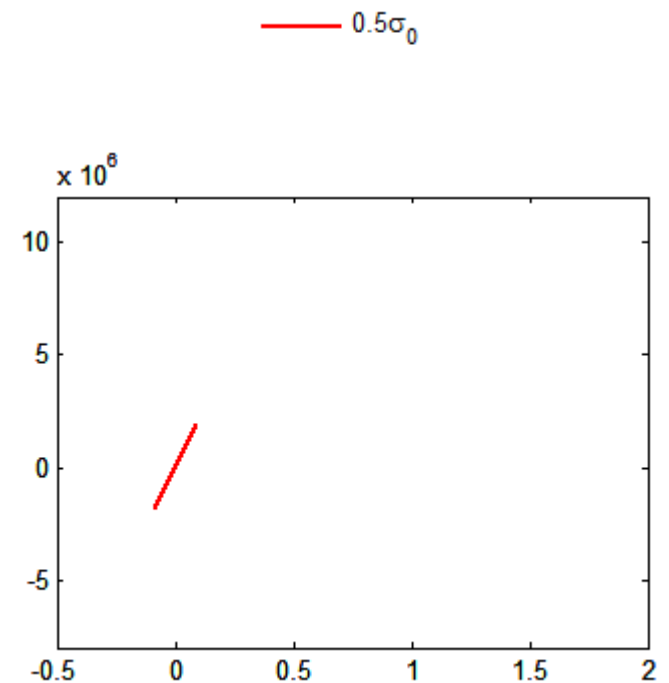
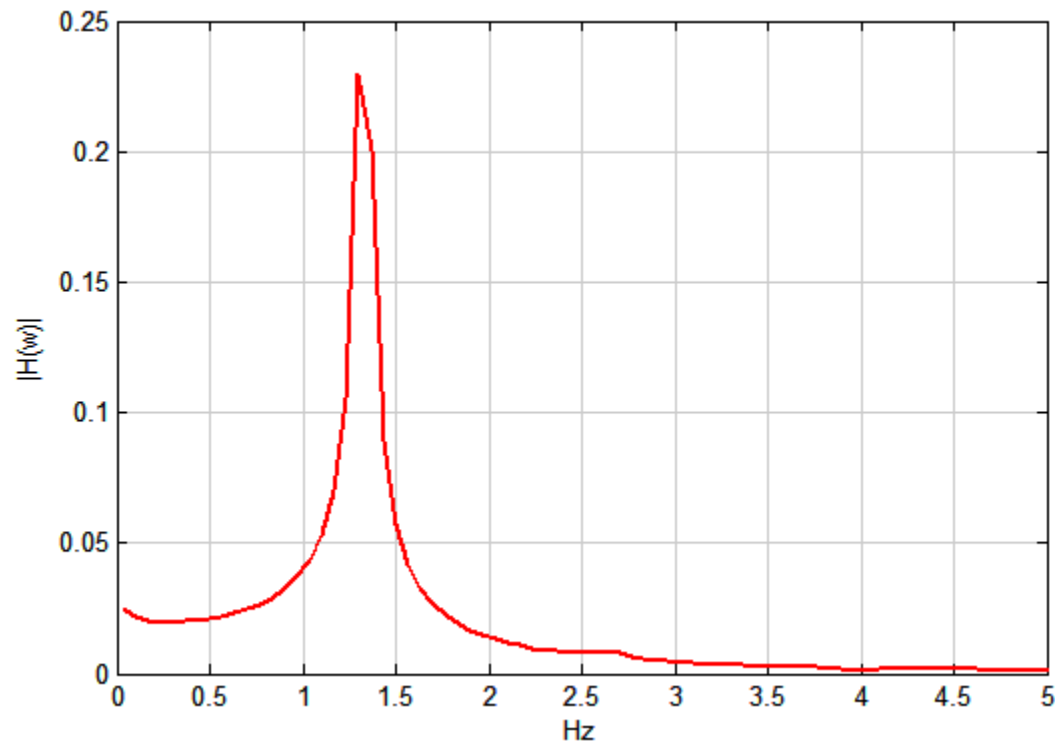
# Characteristics of TELS

27

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FRF of the TELS for selected threshold



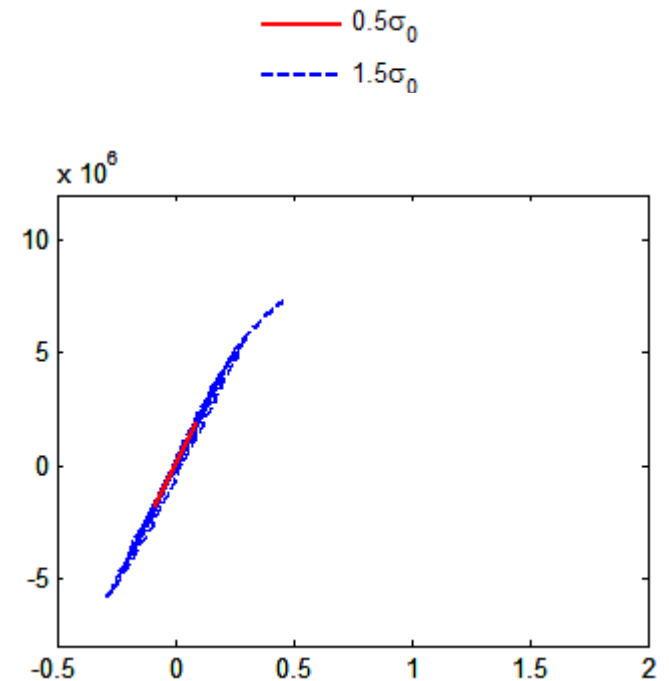
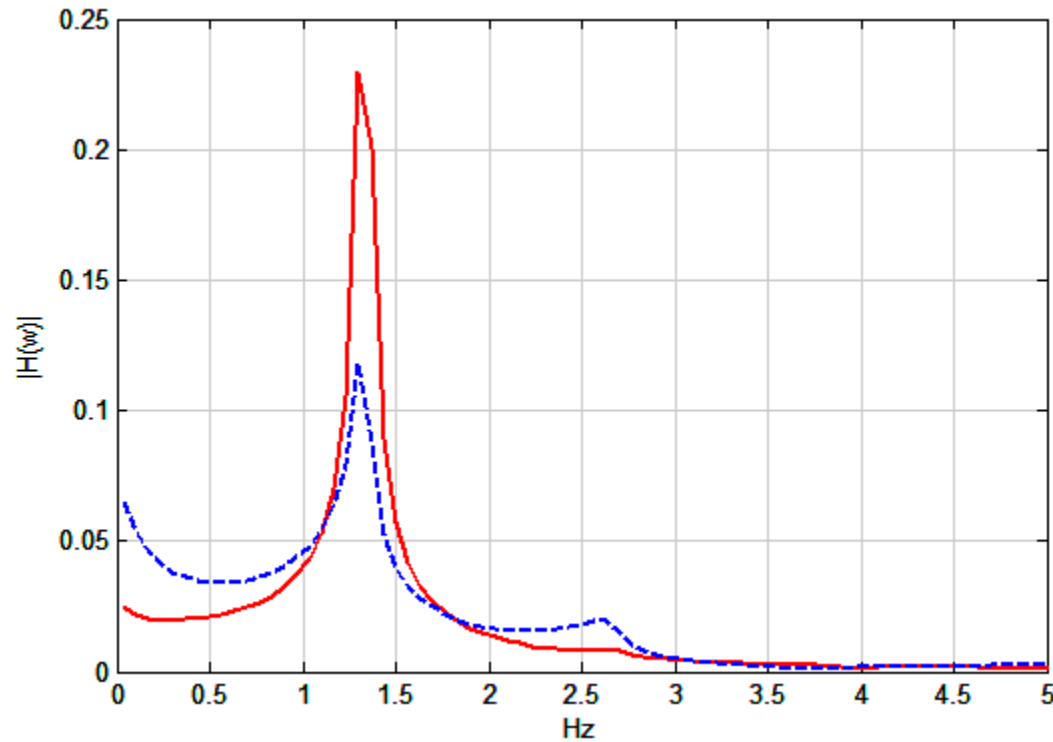
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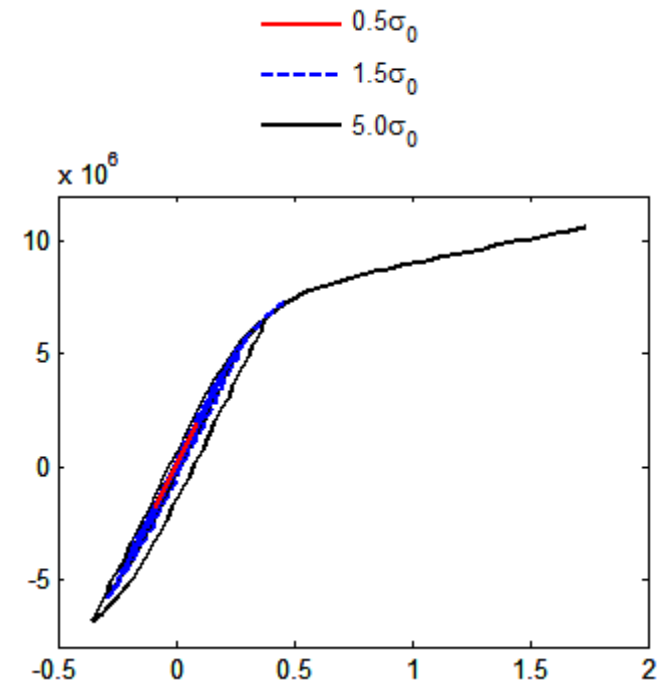
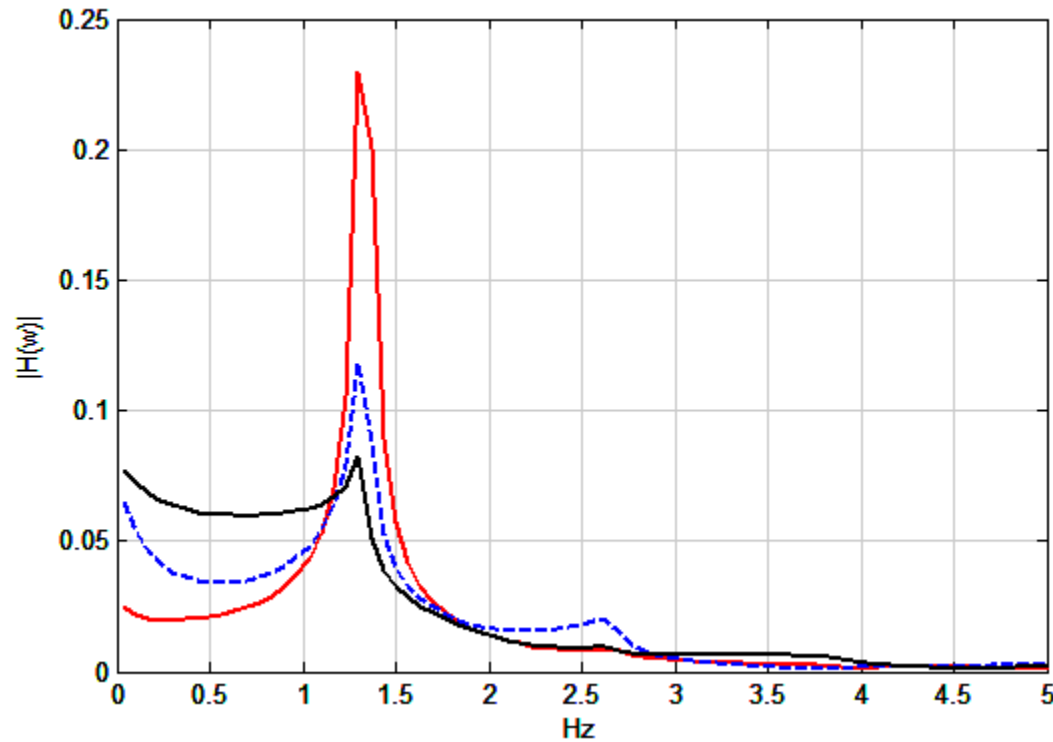
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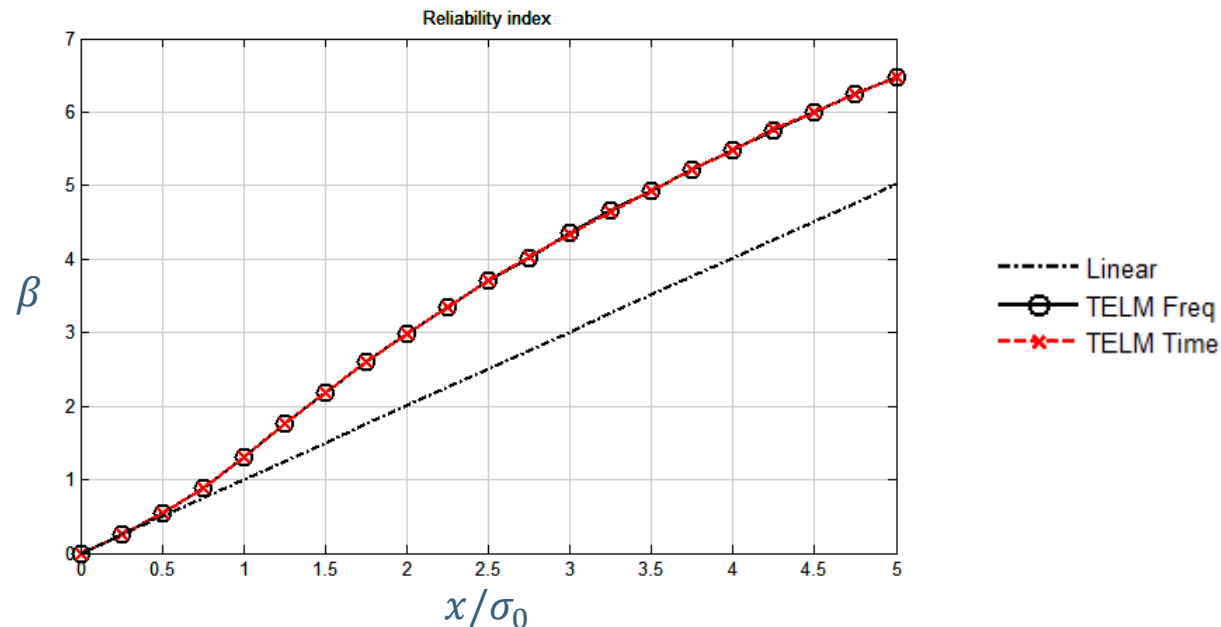


# Characteristics of TELS

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- Because of TELS's dependence on the threshold, TELM captures the *non-Gaussian distribution* of the nonlinear response.

$$\Pr[x \leq X(t, u)] \approx \Phi[-\beta(x)]$$

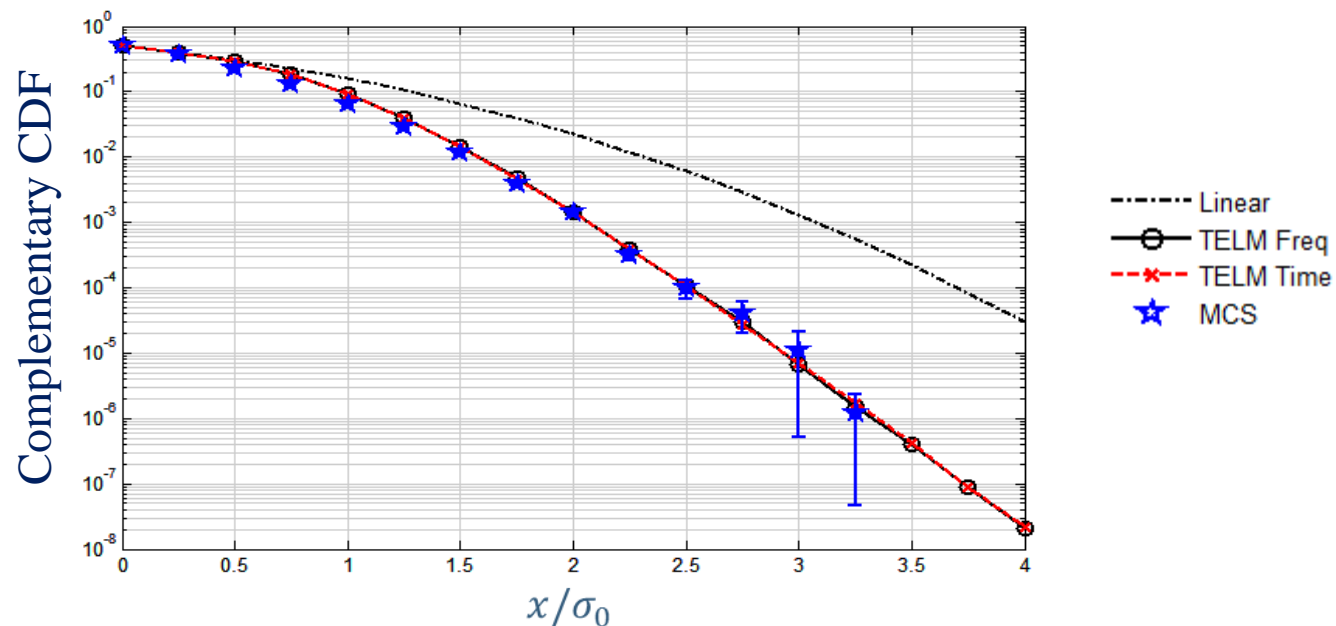


# Characteristics of TELS

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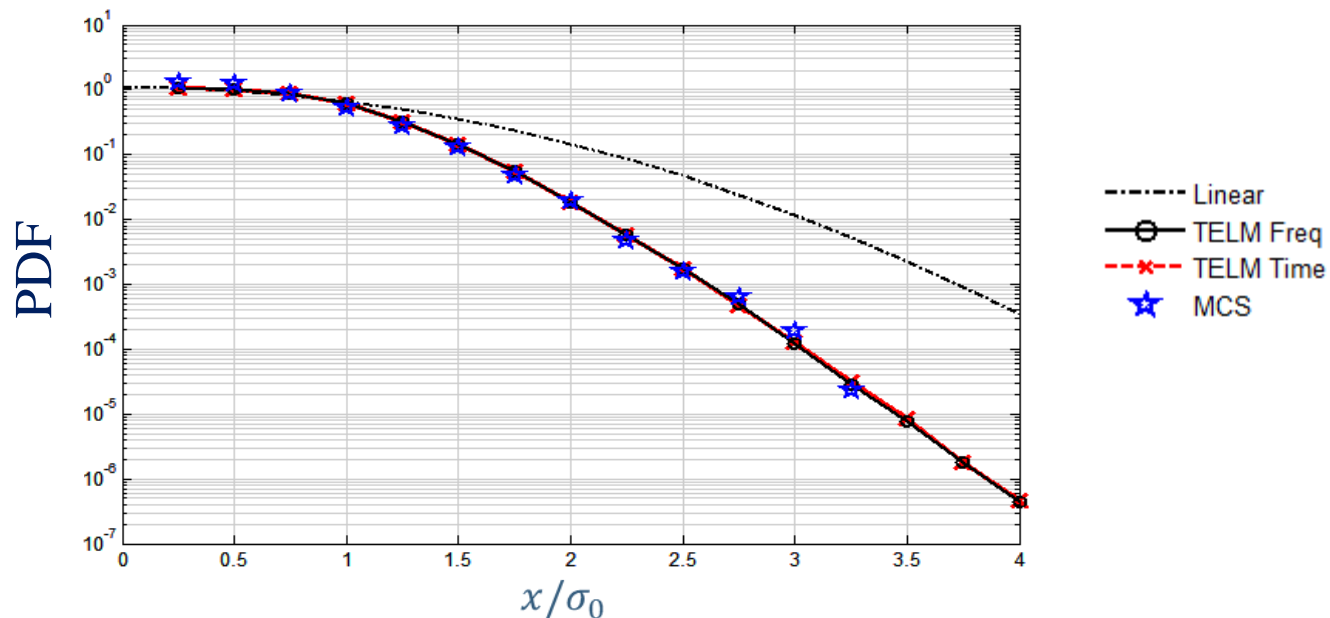


# Characteristics of TELM and TELS

■32

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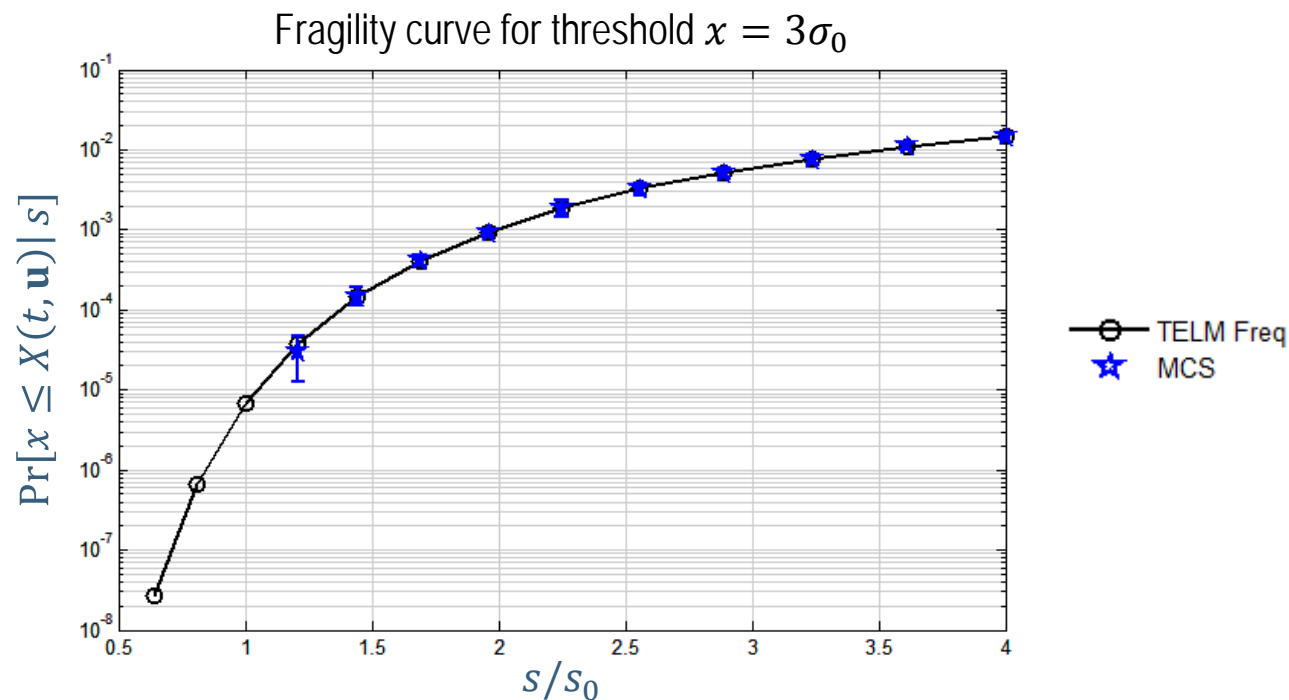
# Characteristics of TELS

■33

- TELS is invariant of the scaling of the excitation, i.e.,  $h(t, x)$  and  $H(\omega, x)$  for excitation  $sF(t)$  are invariant of  $s$ .



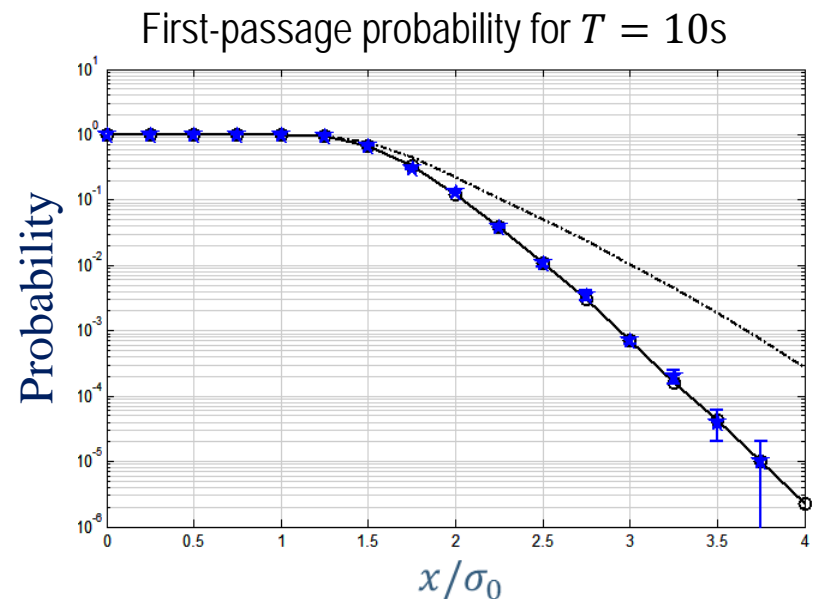
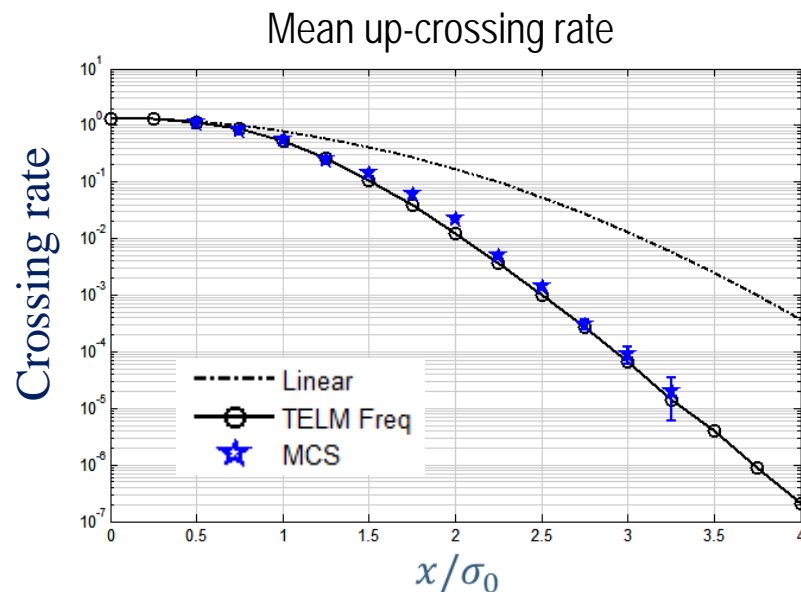
Useful property for developing *fragility curves*:



# Characteristics of TELS

■34

- For stationary response, TELS is invariant of time  $t$ . Thus, TELSs determined for one time point are sufficient to evaluate all statistical properties of the response, e.g.,
  - *Point-in-time distribution*  $\Pr[x \leq X(t, \mathbf{u})]$
  - *Mean up-crossing rate*  $\nu^+(x)$
  - *First-passage probability*  $\Pr[x \leq \max_{0 \leq t \leq T} |X(t)|]$



# Characteristics of TELS

■35

- For non-stationary excitation with smooth evolution in time, TELS at a critical time can be used as an approximation for all times.

➡ *Evolutionary TELM* (Broccardo/Der Kiureghian 2013)

- Evolutionary input-output for a linear system

$$\Phi_{XX}(\omega, t) = |M(\omega, t)|^2 \Phi_{FF}(\omega)$$

$$M(\omega, t) = \int_0^t A(\omega, t - \tau) h(\tau) e^{-i\omega\tau} d\tau$$

- ETELM

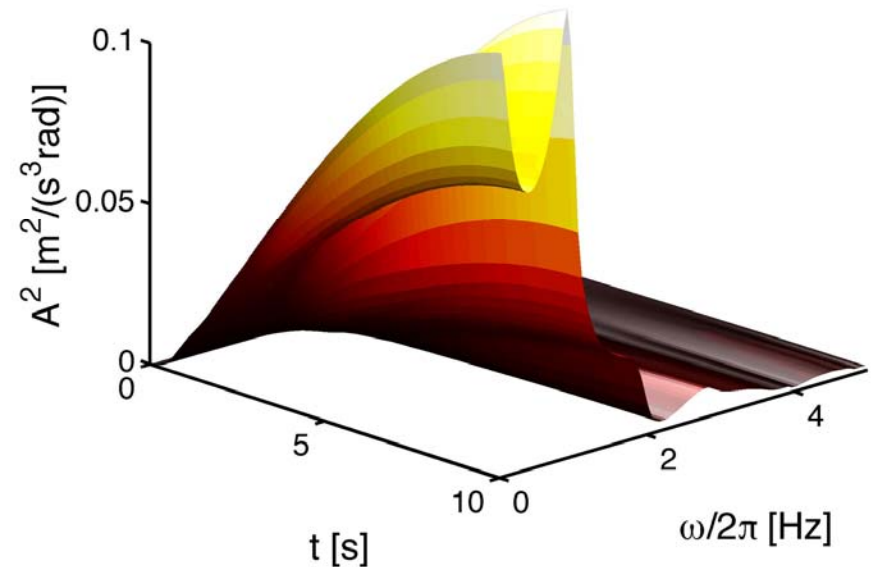
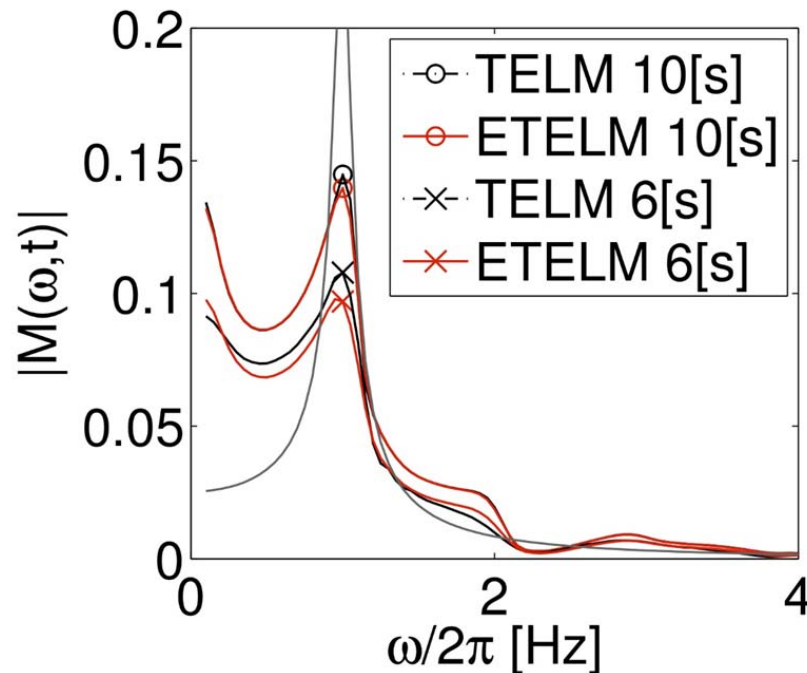
$$M(\omega, t, x) \approx \int_0^t A(\omega, t - \tau) h_{TELS}(\tau, x) e^{-i\omega\tau} d\tau$$

$h_{TELS}(t, x)$  determined at a critical point in time, e.g., peak intensity of the excitation.

# Characteristics of TELS

■36

- For non-stationary excitation with smooth evolution in time, TELS at a critical time can be used as an approximation for all times.
- Example – response to uniformly modulated process

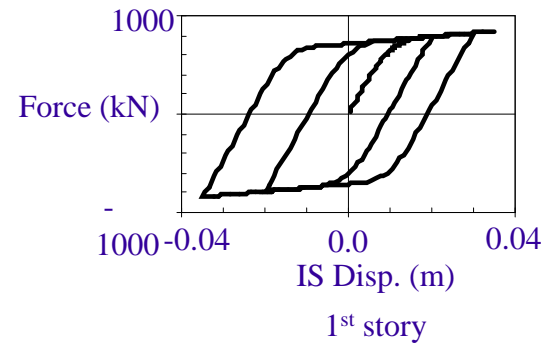
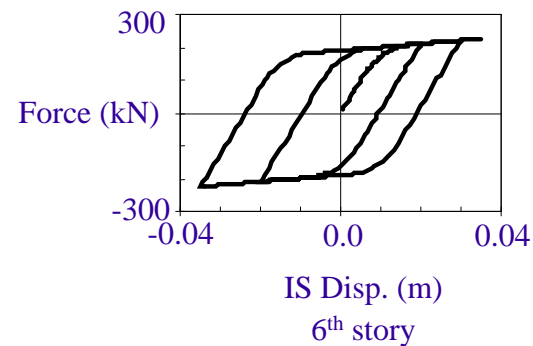
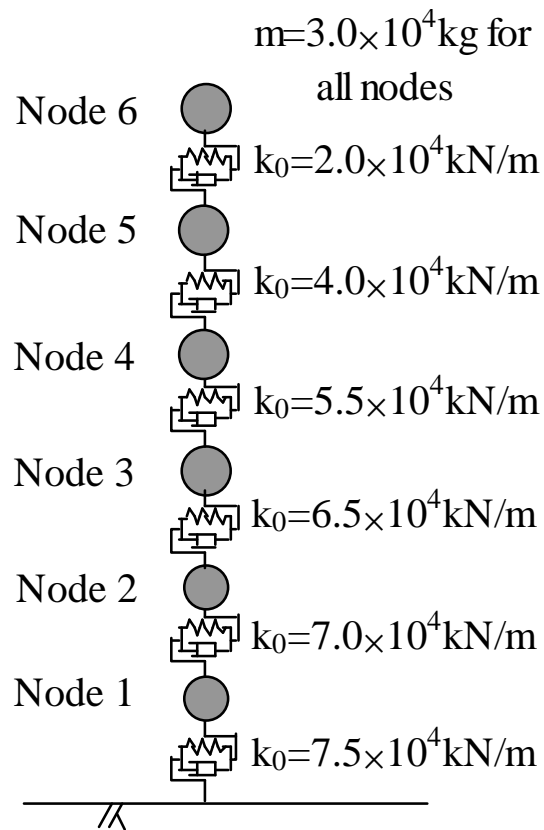




# Applications of TELM

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- TELM is easily extended to *MDoF systems* – number of random variables remain the same.

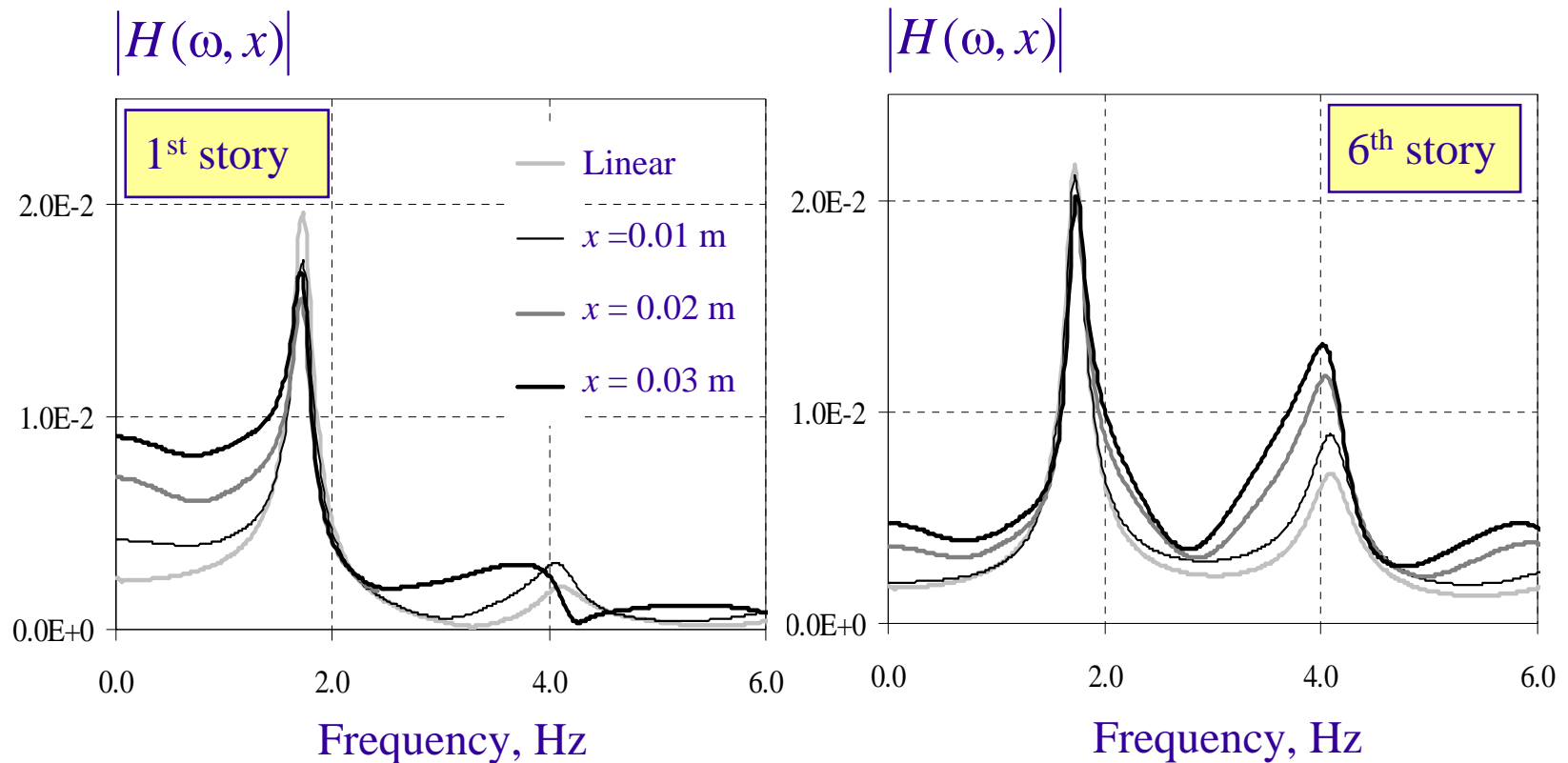


Smooth bilinear hysteresis model

# Applications of TELM

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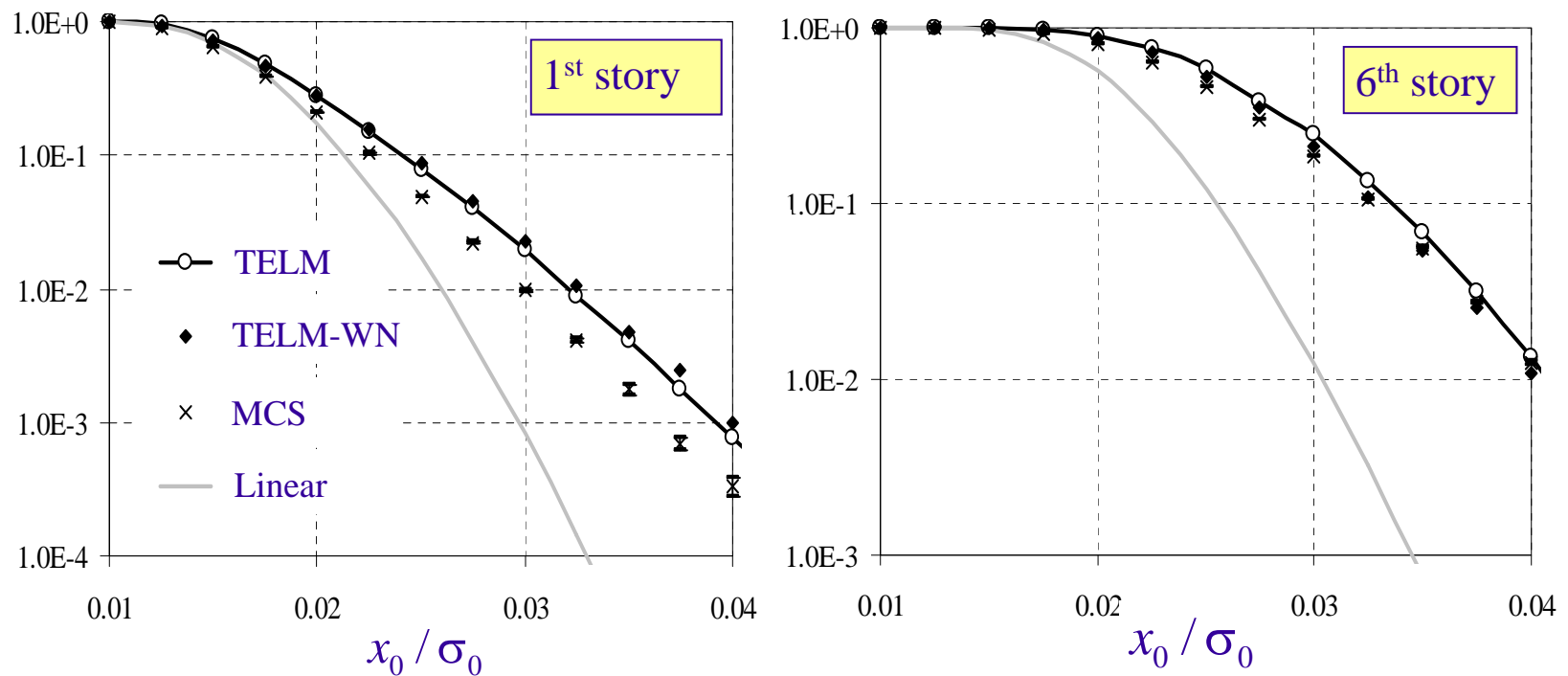
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# Applications of TELM

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First-passage probability (10s duration)

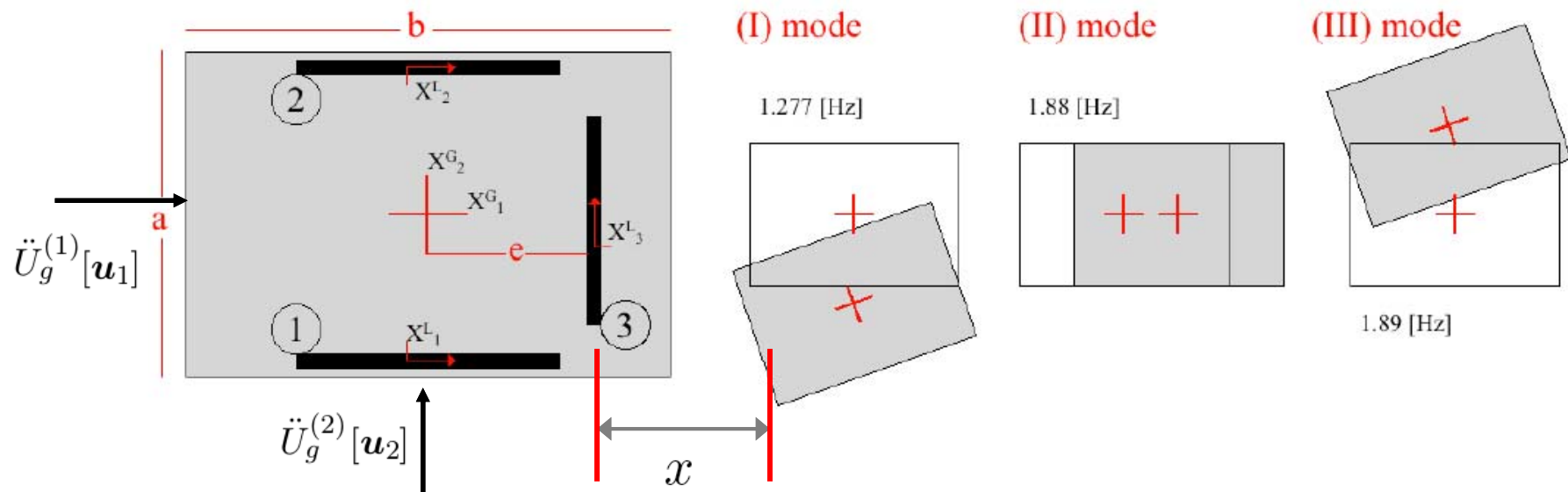
# Applications of TELM

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- Multi-component (SI) excitations

$$\mathbf{F}(t) = \begin{bmatrix} \mathbf{s}^{(1)}(t)\mathbf{u}^{(1)} \\ \vdots \\ \mathbf{s}^{(m)}(t)\mathbf{u}^{(m)} \end{bmatrix}$$

a *separate TELS* is identified for each input component.



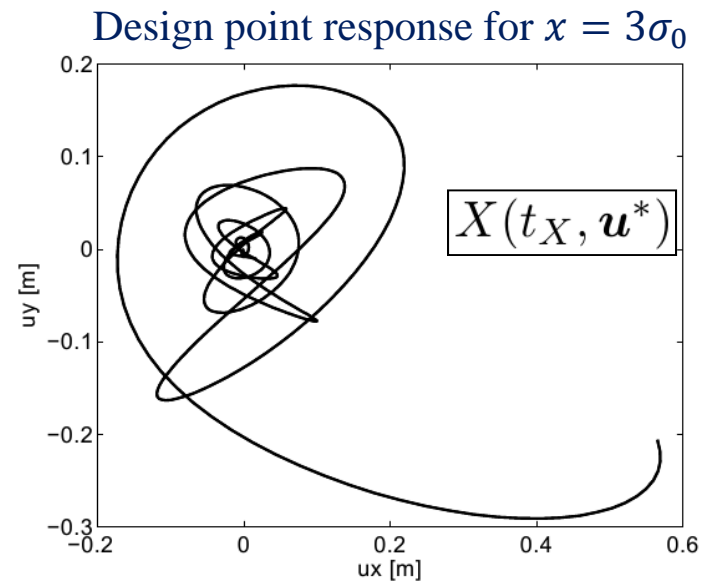
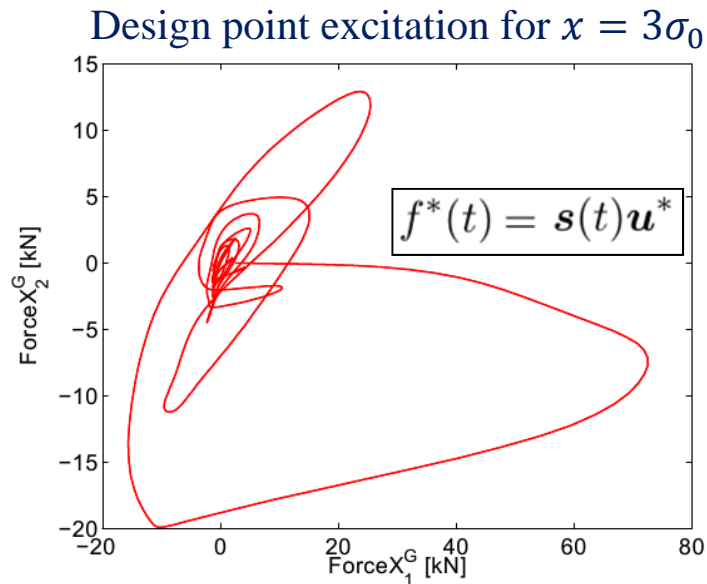
# Applications of TELM

41

## Multi-component (SI) excitations

$$\mathbf{F}(t) = \begin{bmatrix} \mathbf{s}^{(1)}(t)\mathbf{u}^{(1)} \\ \vdots \\ \mathbf{s}^{(m)}(t)\mathbf{u}^{(m)} \end{bmatrix}$$

a *separate TELS* is identified for each input component.



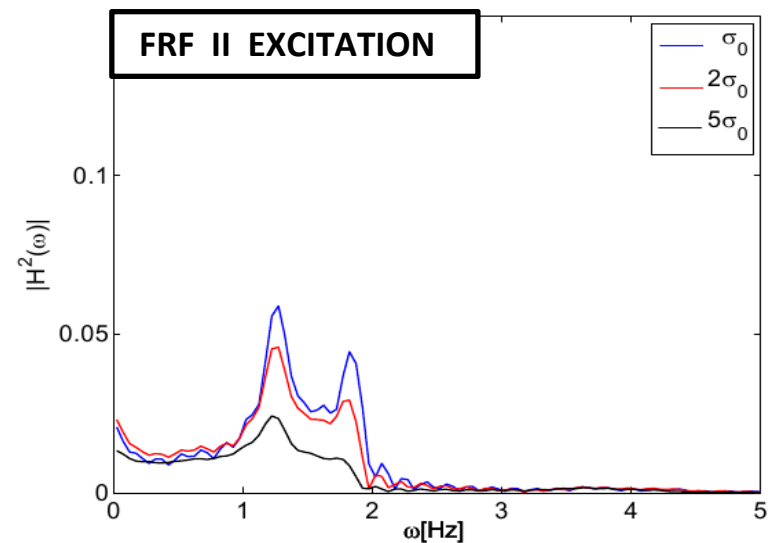
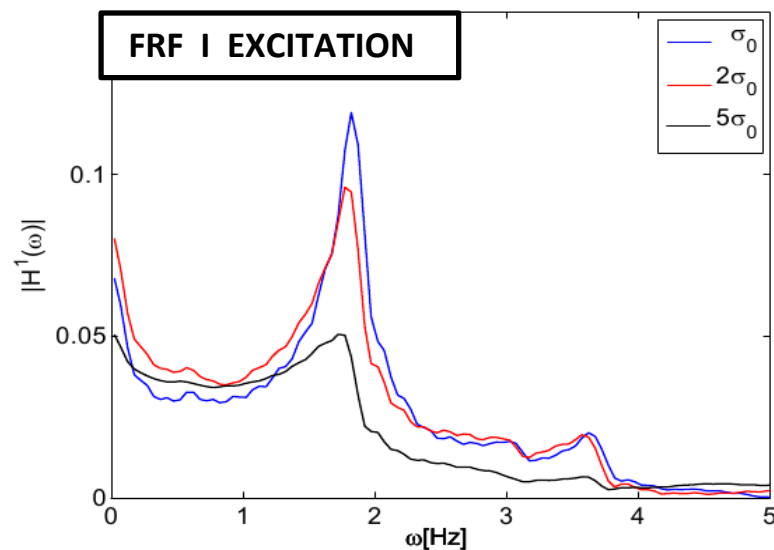
# Applications of TELM

42

## Multi-component (SI) excitations

$$\mathbf{F}(t) = \begin{bmatrix} \mathbf{s}^{(1)}(t)\mathbf{u}^{(1)} \\ \vdots \\ \mathbf{s}^{(m)}(t)\mathbf{u}^{(m)} \end{bmatrix}$$

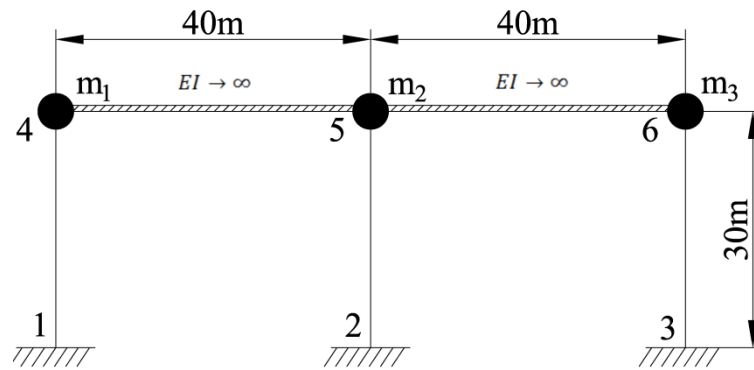
a *separate TELS* is identified for each input component.



# Applications of TELM

■43

- Multiply-supported inelastic system subject to spatially varying ground motions:



$$D_k(t) = \sum_{p=1}^{n/2} [A_{pk} \cos(\omega_p t) + B_{pk} \sin(\omega_p t)], \quad k = 1, \dots, m$$

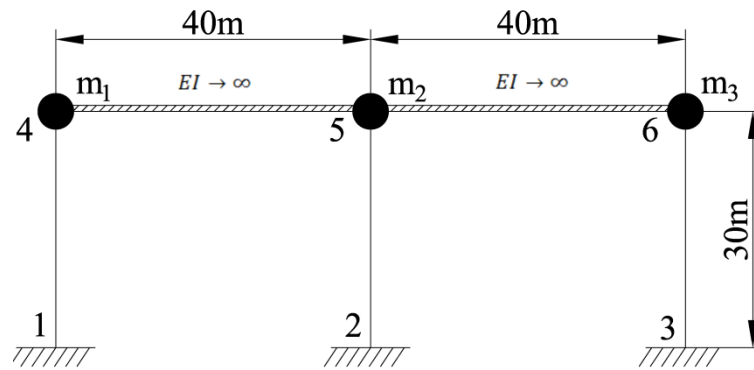
$m$  = number of support DOFs

$A_{pk}, B_{pk}$  = Fourier coefficients, zero-mean Gaussian random variables, independent for different frequencies, correlated for same frequency different supports

# Applications of TELM

■44

- Multiply-supported inelastic system subject to spatially varying ground motions:



Cases considered:

1. Uniform ground motions
2. Totally incoherent ground motions
3. Wave-passage and partial incoherence
4. Case 3 with variable soil conditions

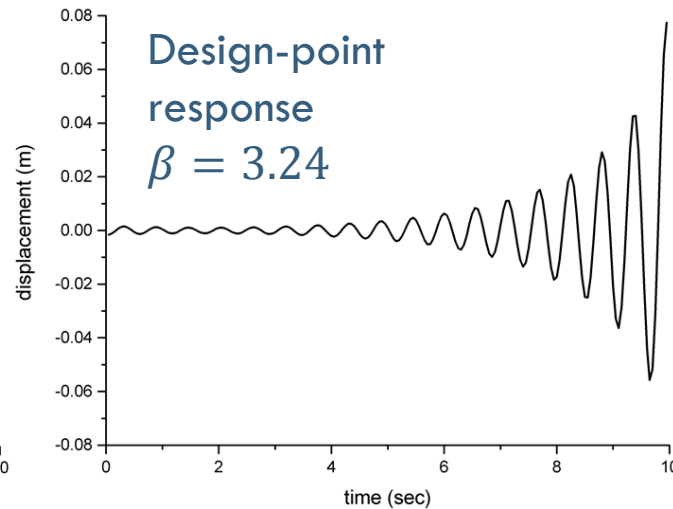
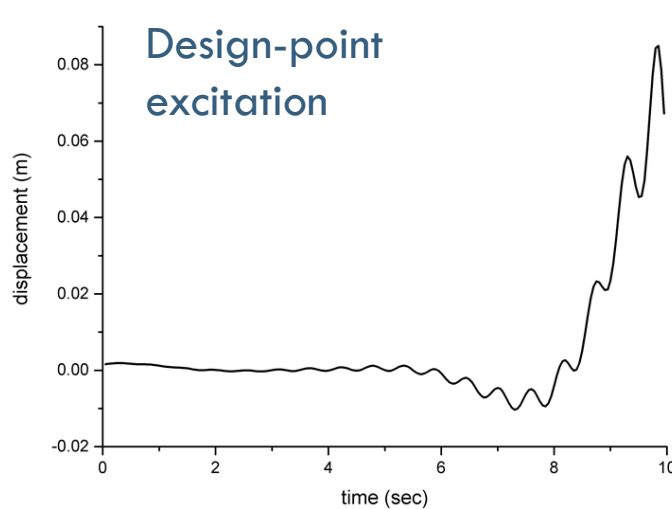
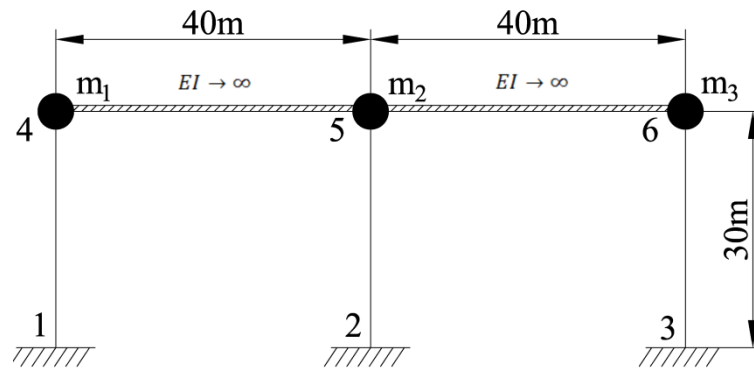


# Applications of TELM

45

- Multiply-supported inelastic system subject to spatially varying ground motions:

Uniform ground motions

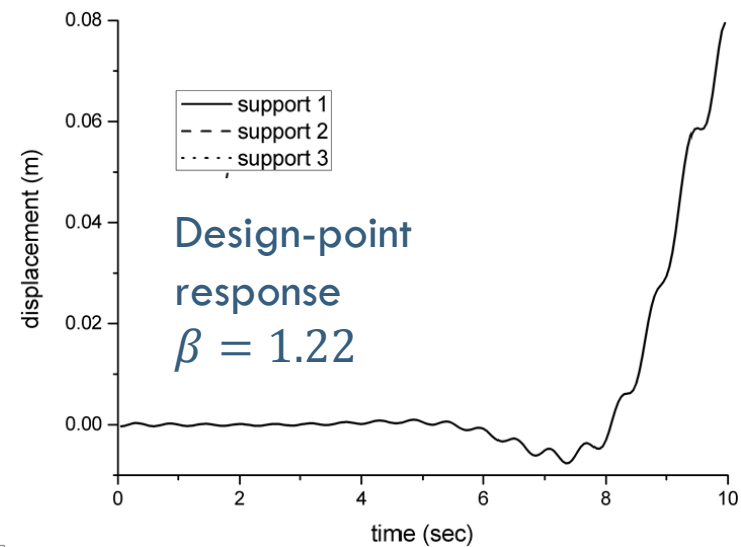
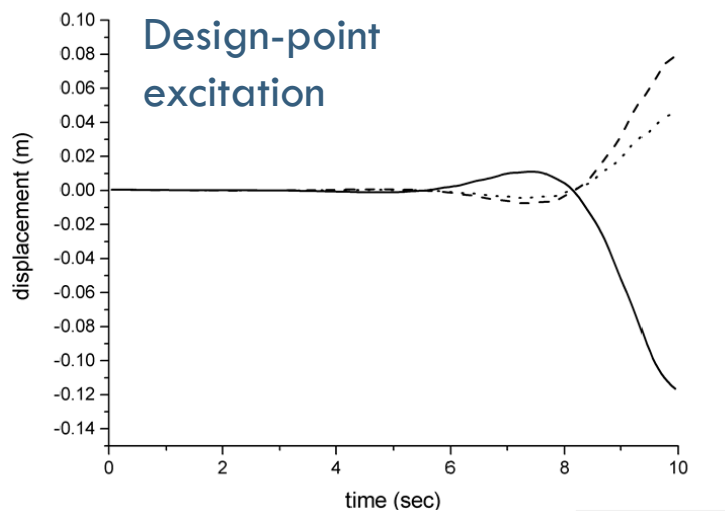
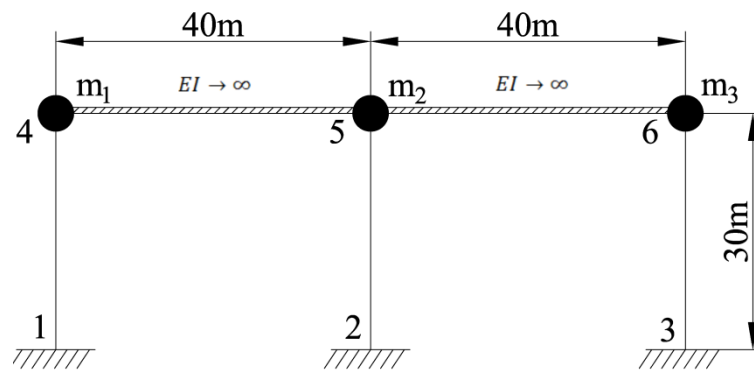


# Applications of TELM

46

- Multiply-supported inelastic system subject to spatially varying ground motions:

Totally incoherent ground motions

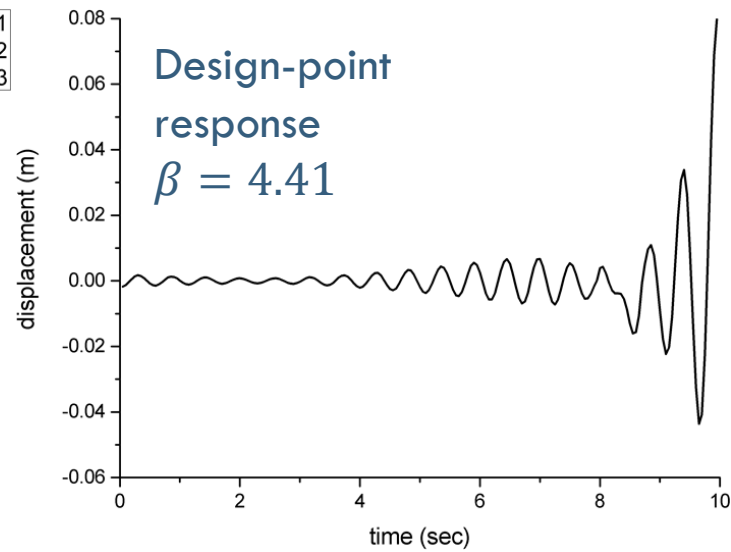
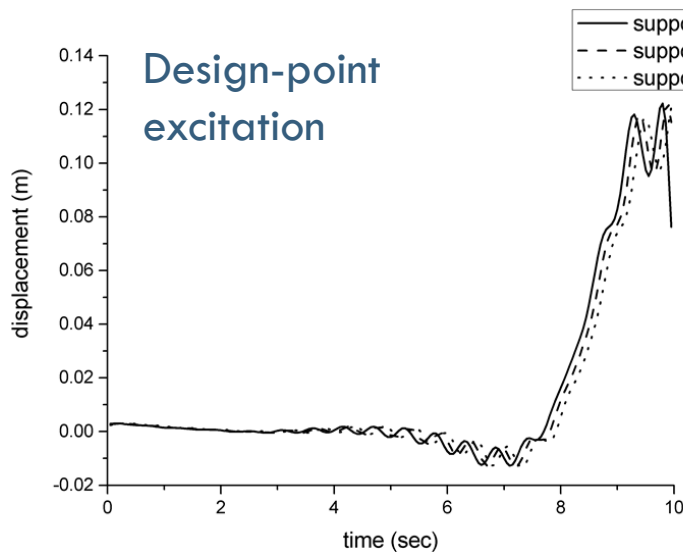
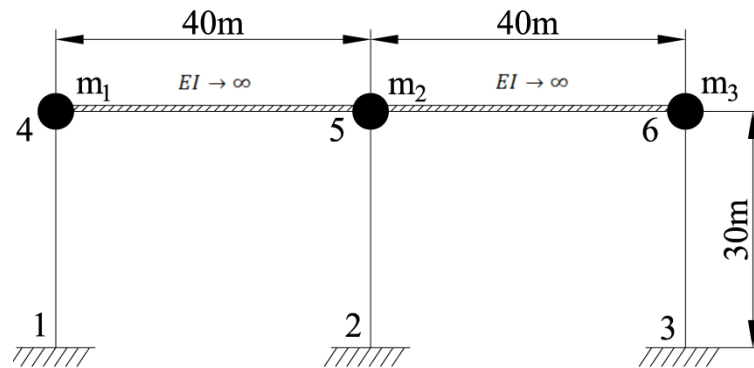


# Applications of TELM

47

- Multiply-supported inelastic system subject to spatially varying ground motions:

Wave passage and partial incoherence

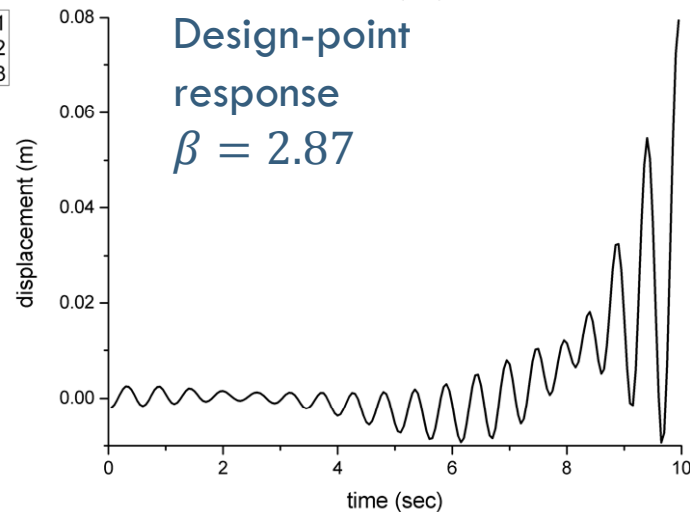
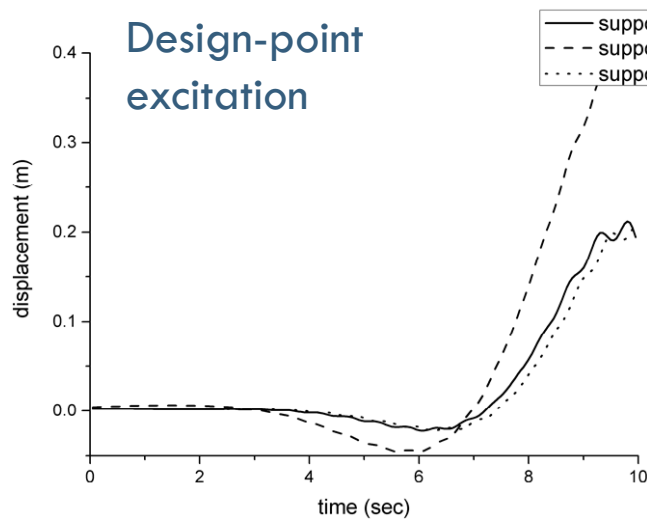
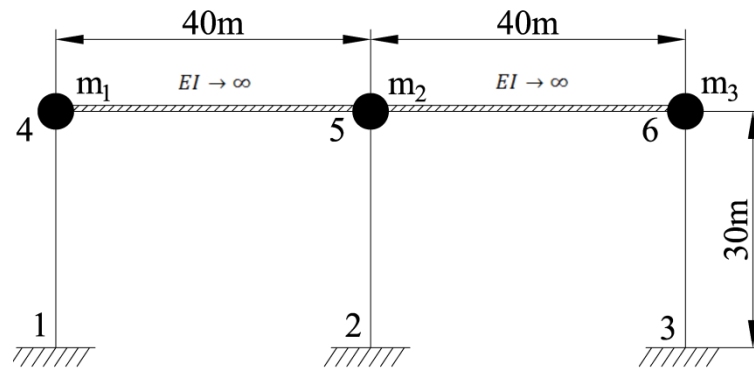


# Applications of TELM

■48

- Multiply-supported inelastic system subject to spatially varying ground motions:

Wave passage and  
partial incoherence  
with variable soil

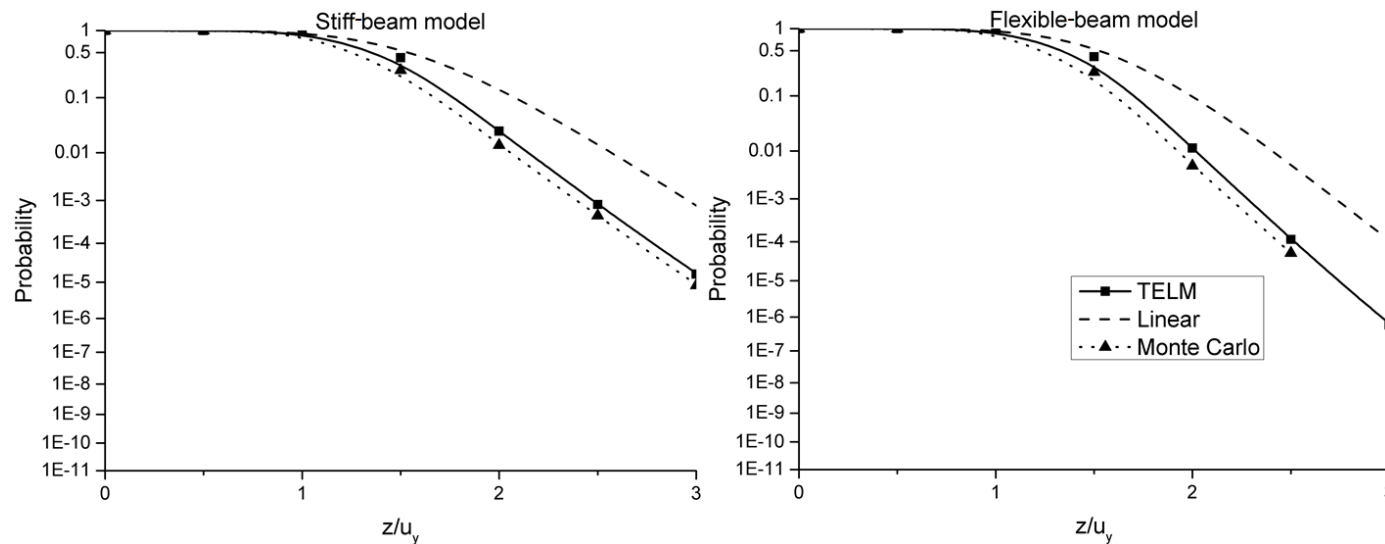
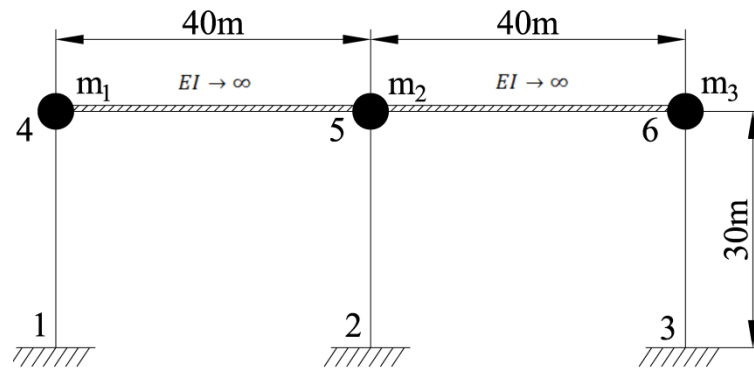


# Applications of TELM

49

- Multiply-supported inelastic system subject to spatially varying ground motions:

First-passage  
probability for  
Case 3



# Challenges and limitations of TELM

■50

- ❑ TELM requires repeated computations of  $X(t, \mathbf{u})$  and  $\nabla_{\mathbf{u}}X(t, \mathbf{u})$  for selected values of  $\mathbf{u}$  (typically around 10 times) to find the design point  $\mathbf{u}^*$ . We use the Direct Differentiation Method (DDM) for this purpose.
- ❑ The nonlinear response must be continuously differentiable – must use smooth or smoothened constitutive laws.
- ❑ The limit-state surface must be well behaving. TELM does not work well for strongly stiffening systems (e.g., Duffing oscillator with a strong cubic term) or when nonlinearity involves abrupt changes in the system behavior.
- ❑ As of now, TELM is not applicable to degrading systems.

# Concluding remarks

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- ❑ TELM is an alternative equivalent linearization method for nonlinear stochastic dynamic analysis.
- ❑ TELM: Is a non-parametric method
  - Captures non-Gaussian distribution of nonlinear response
  - Offers superior accuracy for tail probabilities
  - Is particularly convenient for fragility analysis
  - Can be applied to stationary or non-stationary response
  - Can be applied to MDoF systems, multi-component excitations, variable support motions
- ❑ TELM requires continuous differentiability of the nonlinear response.
- ❑ As with other linearization methods, the accuracy of TELM depends on the nature of the nonlinearity.

Thank you!



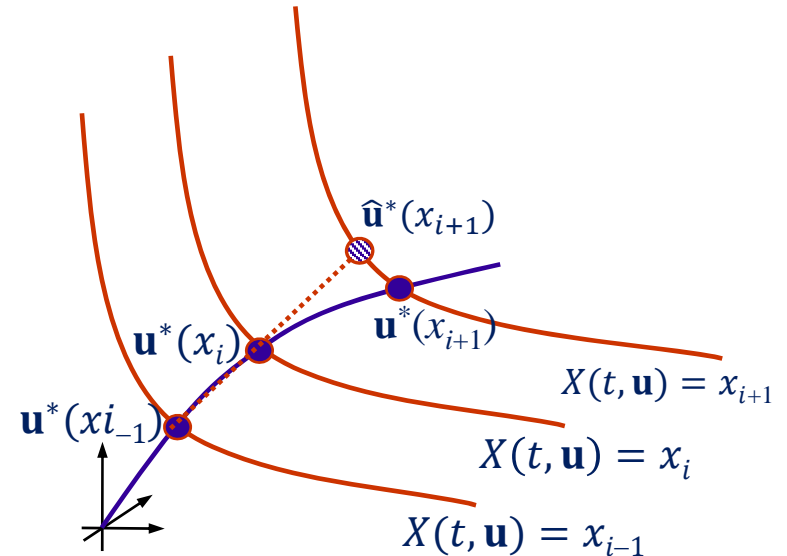
# Determination of the design point

■53

- Iterative algorithms for solving
$$\mathbf{u}^*(x) = \arg \min \{ \|\mathbf{u}\| \mid G(\mathbf{u}, x) = 0 \}$$
require repeated computations of  $X(t, \mathbf{u})$  and  $\nabla_{\mathbf{u}} X(t, \mathbf{u})$ .

- For an ordered sequence of thresholds  $x_1 < x_2 < \dots < x_n$ , use extrapolated starting points:

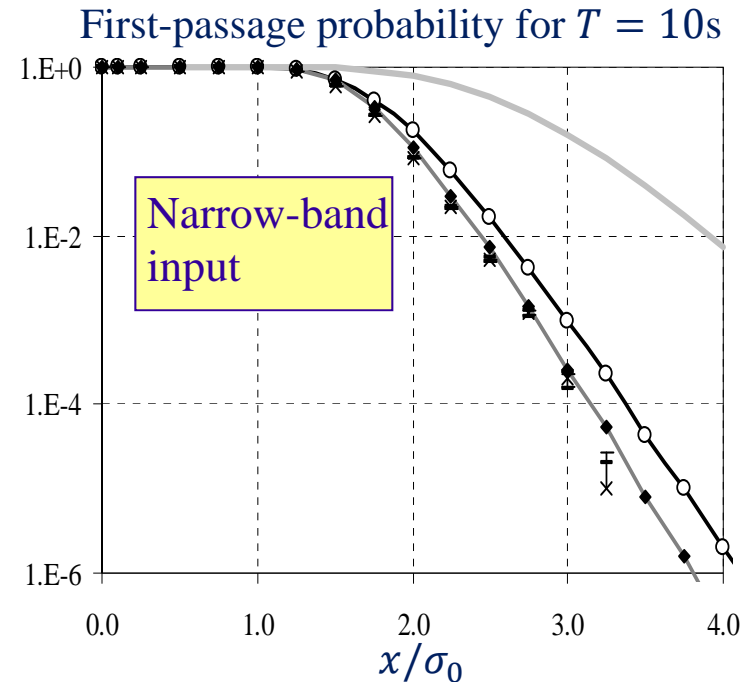
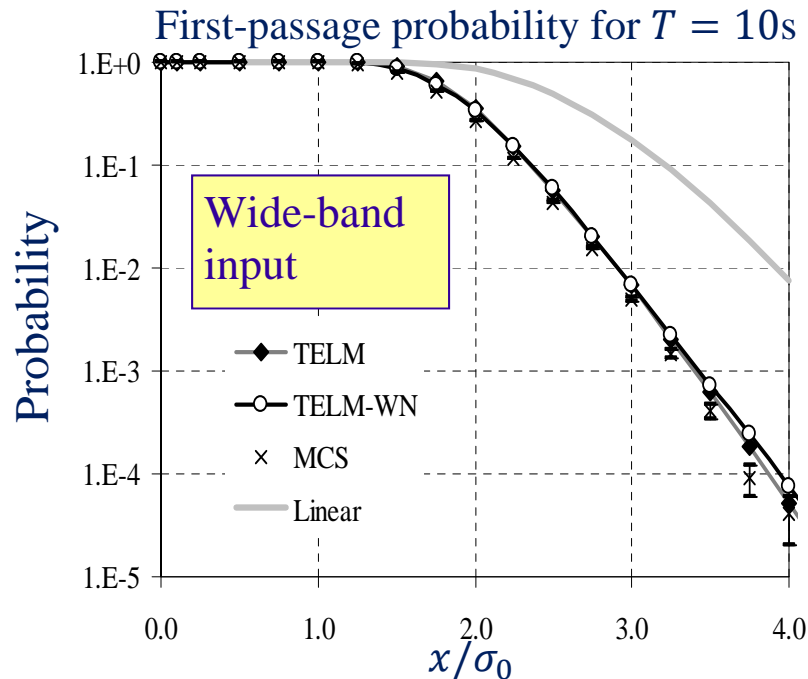
$$\hat{\mathbf{u}}^*(x_{i+1}) = \mathbf{u}^*(x_i) + \lambda \frac{\mathbf{u}^*(x_i) - \mathbf{u}^*(x_{i-1})}{\|\mathbf{u}^*(x_i) - \mathbf{u}^*(x_{i-1})\|}$$



# Characteristics of TELM and TELS

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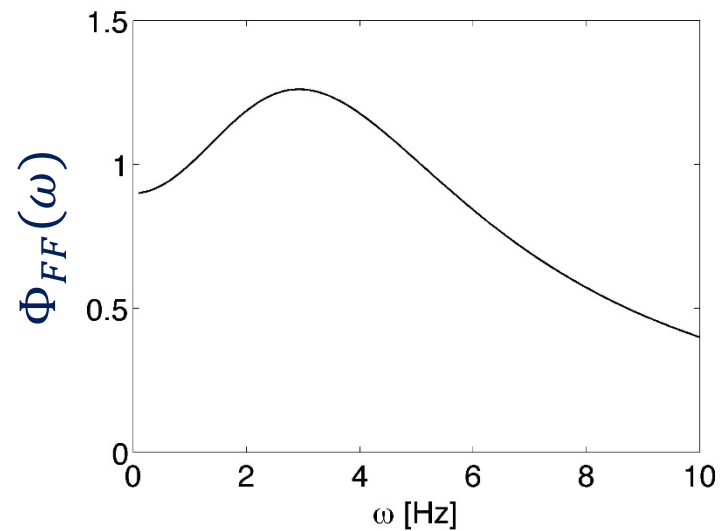
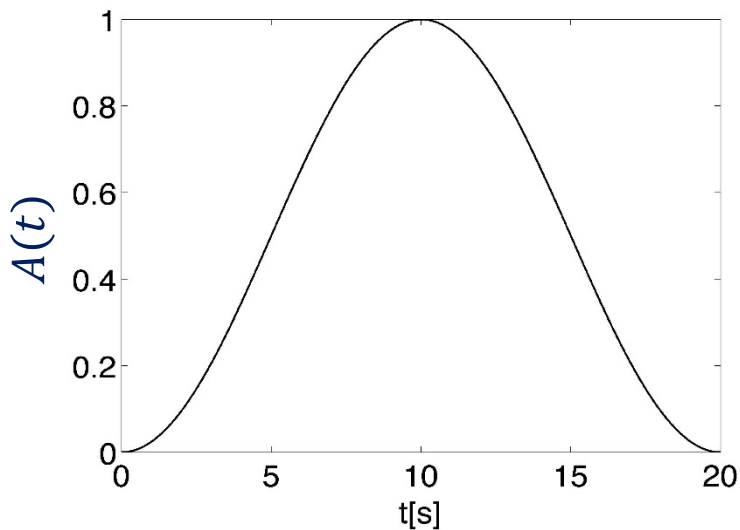
- For broad-band excitations, the TELS is insensitive to the frequency content.
- ➔ TELS for response to white noise can be used as an approximation for response to *non-white excitations*



# Characteristics of TELM and TELS

■55

- For non-stationary processes with smooth evolution in time, TELS at a critical time can be used as an approximation for all times.
- Example – response to uniformly modulated process



# Characteristics of TELM and TELS

56

- For non-stationary processes with smooth evolution in time, TELS at a critical time can be used as an approximation for all times.
- Example – response to uniformly modulated process

